# TRP 5000 <br> INSTRUCTION MANUAL 

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## 1. Introduction

1.1 The TRP 5000 is a single side band radiotelephone transmitter receiver combination for duplex, semi-duplex and simplex telephone traffic in the $1.6-4 \mathrm{MHz}$ coastal telephone band and in the maritime mobile short wave bands between 4 and 27.5 MHz . In addition the TRP 5000 provides facilities for radiotelegraph and telex transmission and reception

The TRP 5000 is composed of the T5000transmitter power amplifier, the $P$ $5000(24 \mathrm{~V}$ DC) or the P 5001 (AC) power pack, the E 5000 exciter and the R 5000 receiver. This instruction manual describes the Transmitter Power Amplifier, the Exciter and the Power Packs, while the Receiver is covered in a separate manual.

The transmitter is fully synthesised and provides keyboard selection of up to 256 pre-programmed frequencies, which can be freely distributed in the maritime frequency bands.
Silicon transistors and integrated circuits are used throughout except for the transmitter power amplifier stage. These features in conjunction with the fact that no crystal ovens are used enable the transmitter to be ready for operation within 30 seconds after being switched on.
The cabinet is a sturdy 19 -inch rack construction containing all
necessary inter- connections.

Because we are constant processing the experience gained during the production and operation of our equipment, it is possible for minor modifications to occur relative to the information given in this instruction manual. Wherever practicable, however, any corrections will be listed on a correction sheet at the back of the front cover of this manual.


## 2. Operating Instructions

### 2.1 Distress Operation on, 2182 kHz

Set controls as fellows:

A SUPPLY to TRANSMIT
B C BANDs to 2182 kHz

D VOLUME clockwise

E SENSITIVITY fully clockwise
F MODE to TRANSMIT ALARM

G Press and release ALARM START

The alarm signal is now transmitted for approx. 45 seconds and may be monitored in the handset earpiece. When the alarm signal ceases depress handset key and, speaking clearly into microphone, transmit distress message.

If it is required to repeat the alarm signal transmission, it Is only necessary to press and release the ALARM START push button again.

An alarm signal transmission may be interrupted at any time by turning the MODE switch to A3H.

NOTE: In the TRANSMIT ALARM mode on 2182 kHz the power output of the transmitter is automatically set to FULL POWER SIMPLEX independent of the setting of the POWER switch.

### 2.2 Tuning to 2182 kHz

1. Set SUPPLY switch to TRANSMIT.

30 seconds after switching on the transmitter is ready for operation.
2. Set BAND switch to 2182 kHz .

The band-indicator lamp will show constant light indicatingthat2182
kHz is selected. The FREQUENCY NO display will extinguish.
3. Press TUNE button and adjust TUNING control for maximum deflection on ANTENNA CURRENT meter.
The tuning range on 2182 kHz is reduced so that power is transmitted even when the TUNIN-G control is not adjusted.

The transmitter is now ready for operation.

NOTE: The type of service used on 2182 kHz is A3H, simplex. This mode is automatically selected when the BAND switch is set to 2182 kHz , and the MODE switch can therefore be set to any position except TESTALARM. The equipment will also work in the simplex mode even if the POWER switch is in a DUPLEX position.

### 2.3 Tuning to a Frequency Other than 2182 kHz

1. Set SUPPLY switch to TRANSMIT. 30 seconds after switching on the transmitter is ready for operation,
2. Set BAND switch to any band except 2182 kHz
3. Turn DIMMER control fully clockwise
4. If FREQUENCY NO display does not show zero, clear display using C key of keyboard.
5. Look up desired frequency in frequency chart and read Frequency -No
6. Key Frequency No. into keyboard. The selected Frequency No. will be displayed.
7. Set BAND switch to position indicated by flashing band indicator lamp. If no flashing occurs the BAND switch is already correctly set.
8. Press TUNE button and adjust TUNING control for maximum deflection on ANTENNA CURRENT meter.

9 Select desired type of service with MODE and POWER switches.

Transmitter is now ready for operation

NOTE: Transmission is inhibited if the mode setting does not correspond with the frequency selected. On radiotelephone frequencies the transmitter can be keyed only in the modes A3A, A3J, A3H, and TRANSMIT ALARM. On wireless telegram frequencies the transmitter can be keyed only in the modes Al, A2H, and Fl.

### 2.4 Operating Controls and their Functions

2.4.1 The SUPPLY switch has four positions

OFF Receiver and transmitter are switched off.

REGEIVE ONLY Power Pack is started up and supplies power to Receiver (and grid bias to transmitter output valves). Remote speaker of receiver is connected to receiver output.

STAND BY Power Pack supplies power to Receiver, Exciter, band indicator, and filaments of transmitter output valves. Remote speaker of receiver is connected to receiver output.

TRANSMIT Transmitter can be keyed. Remote speaker of receiver is disconnected.

NOTE:A built-in delay circuit protects the output valves of the transmitter from being keyed for the first 30 sec . after switching to STAND BY or TRANSMIT.

LOW POWER SIMPLEX

MEDIUM POWER SIMPLEX

FULL POWER SIMPLEX

Transmitter is keyed from handset key, morse key, or telex equipment dependent on mode of operation. Receiver is muted while transmitting.
Transmitter can be driven to approx. 1/20 Of full output power.

As above, but transmitter can be driven to approx. 1/4 of full output power.

As above, but transmitter can be driven to full power.

FULL POWER DUPLEX

MEDIUM POWER DUPLEX

LOW POWER DUPLEX

Transmitter is keyed constantly in the Fl, A3A, A3J, A3H and TRANSMIT ALARM modes (provided the selected frequency is in accordance with the mode setting). Receiver is on but built-in speaker is disconnected. In the $A 2 H$ and $A l$ modes the operation is simplex independent of the setting of the POWER switch. Transmitter can be driven to full output power.

As above, but transmitter can be driven to approx. 1/4 of full output power.

As above, but transmitter can be driven to approx. 1/20 of full output power.

NOTE: With the BAND switch set to 2182 kHz , simplex is automatically selected, independent of the POWER switch setting, and if at the same time the MODE switch is set to TRANSMITALARM, the transmitter is automatically set to full power.

The MODE switch has eight positions

A2H Transmission of modulated radiotelegraphy. Only the morse key input is open. The transmitter can only be keyed if a telegraphy frequency is selected.
Al Transmission of un-modulated radiotelegraphy. Only the morse key input is open. The transmitter can only be keyed if a telegraphy frequency is selected.
F1 Transmission of telex. Only the telex inputs are open. Trans- mission is only possible if a telegraphy frequency is selected.
TEST ALARM The built-in two tone alarm generator is connected to the receiver AF amplifier. Transmitter cannot be keyed. A3A Transmission of single-side band signal with reduced carrier. Transmission is only possible if a radio telephony frequency is selected. The transmitter can be keyed from the hand set key or by setting the POWER switch to DUPLEX.
A3J As above, but carrier suppressed
A3H As above, but full carrier.
TRANSMIT ALARM The two-tone alarm generator is connected to the receiver AF amplifier and the alarm generator is ready for transmission of an alarm signal. The mode is A3H as above.
2.4.4 The ALARM START push-button is used to start the alarm generator after the MODE switch has been turned to the TRANSMIT ALARM position. The push-button is depressed and released and the alarm signal will be transmitted for approx. 45 seconds.

The push-button is also used for starting the alarm generator in the TEST ALARM position.
2.4.5 The KEYBOARD is used for frequency selection. It controls the frequency memory. The programmed frequencies are listed in the frequency chart. Each frequency is supplied with a number and this number is keyed into the keyboard. The frequency number selected is displayed on the FREQUENCY NO display. The display is cleared by using the C-key, which resets the display to zero.
The display extinguishes when the BAND switch is set to 2182 kHz , but keyboard entry is still possible and the display will show the selected Frequency No. when the BAND switch is turned away from 2182 kHz .

If an un-programmed Frequency $N o$ is selected, the display will start to flash, indicating that transmission is not possible.
2.4.6 The DIMMER controls the intensity of the FREQUENCY NO display.
2.4.7 The TUNING control is used for tuning the antenna circuit to maximum antenna current indicated on the ANTENNA CURRENT meter.
2.4.8 The BAND switch has 16 positions:

2182 kHz The FREQUENCY NO display is extinguished and the bandindicator lamp shows constant light indicating that 2182 kHz is selected. The range of the TUNING control is reduced. The A3H, simplex mode is automatically selected.

Bands A to H cover the coastal telephone band 1.6 to 4 MHz . The $4 \mathrm{MHz}, 6 \mathrm{MHz}, 8 \mathrm{MHz}, 12 \mathrm{MHz}, 16 \mathrm{MHz}, 22 \mathrm{MHz}$, and 25 MHz positions cover the maritime short-wave bands.

A band indicator lamp at each position shows by flashing light where to set the BAND switch in accordance with the Frequency No selected. The light will extinguish when the BAND switch is set at the correct position.
2.4.9 The TUNE push-button is used when tuning the antenna circuit. The transmitter is keyed and a tune signal is generated. During tuning, the receiver is muted.
2.4.10 The CHECK SWITCH is not normally operative. Pulling the switch knob out will switch the ANTENNA CURRENT meter to read the voltage or current selected with the switch. When released, the knob will return to its original position.

The switch has eight positions:

LEVEL Check of anode AG voltage swing
VA Check of anode DC voltage
Vs1 Check of screen grid voltage of valve no. 1
Vs2 Check of screen grid voltage of valve no. 2
VG Check of grid bias
I1 Check of cathode current of valve no. 1
I2 Check of cathode current of valve no. 2
VD Check of supply voltage to driver amplifier.

### 2.5 Filter Cleaning

The air filter on the transmitter power amplifier front panel should be cleaned periodically. The filter element is easily removed without the use of tools. Wash in warm water and dry thoroughly.

Correct installation of the equipment is important for maximum performance and reliability. Antennas and earth connections must be installed with the greatest care, especially where duplex telephony is desired.

### 3.1 Types of Installation

The TRP 5000 radiotelephone may be powered from either a $24 V$ battery or from 110/115/120/220/230/240V-AC mains.

The TRP 5000 is composed of the following units:

For $24 V$ battery operation:

Type T 5000 transmitter power amplifier
Type P 5000 power pack
Type E 5000 / E5001 exciter
Type R 5000 / R5001 receiver

For AC mains operation (single-phase or two-phase):

Type T 5000 transmitter power amplifier
Type P 5001 power pack
Type E 5000 / E5001 exciter
Type R 5000 / R5001 receiver

The units are connected together in the TRP5000 cabinet in which also the connections to the permanent installation are made.

### 3.2 Removal of units

After removal of the front-panel screws, the units may be pulled out as far as the built-in stops permit. The cables may be unplugged and the individual units removed entirely from the cabinet after having released the stop latches. This is done by pushing the unit slightly back into the cabinet (approx. 10 mm ) and pressing the two nylon knobs of the latches at the sides of the unit while pulling the unit out.

For ease of removal of the units start with the lowest and proceed upwards. The reverse order is used when refitting the units.

### 3.3 Mounting the Cabinet

The cabinet is supplied with shock absorbers. The drawing on page320showsthe drilling plan for the necessary holes.

### 3.4 Connection to the Permanent Installation

Check that the correct power pack is installed in the equipment and, in case of $A C$ operation, set $f$ or the correct mains voltage. The drawing of Terminal strip A on page $3-23$ shows the marking of the terminals to be used for the installation. Necessary cable cross sections are also indicated. All cables except the transmitter antenna and earth leads are brought through the cut out in the cabinet rear wall in a loop that is large enough to take up any play between the equipment and the bulkhead.

A mains switch and fuses are to be provided in the supply leads. External fuse ratings are listed in Table 3.1.

| Power Pack | Supply Voltage |  | External Fuses |
| :---: | :---: | :---: | :---: |
| P 5000 | 24 V | DC | 50A |
| P 5001 | 110 V |  |  |
|  | 115 V 120 V | $\begin{gathered} 50-60 \\ \mathrm{~Hz} \end{gathered}$ | 20A |
| P 5001 | 220 V |  |  |
|  | $230 \mathrm{~V}$ <br> 240 V | $\begin{gathered} 50-60 \\ \mathrm{~Hz} \end{gathered}$ | 15A |

Table 3.1
3.4.1 Supply Voltage Changing in P 5001 (AC operation) :

Voltage Changing in the $P 5001$ power pack is carried out by unsoldering the connections to and between the terminals of the two transformers 265 TI and 265 T 2 .

The necessary connections are shown on the diagram of $P 5001$. When changing voltage it is also necessary to replace the fuse in the power pack input lead. Fuse rating is given on the P 5001 diagram.

### 3.5 Earth Connections

As the transmitter earth connection is always a part of the total antenna system, it is of the utmost importance that the earth connection is constructed to have the smallest possible RF-impedance. Losses in the earth connection will result in a decrease in radiated power which means that the range of the transmitter will be reduced. A poor earth connection will further impede or even make duplex communication impossible.
3.5.1 Transmitter Earth Terminal:

The transmitter earth terminal is located on the top of the cabinet.
3.5.1.1 Steel Ships:

From the transmitter earth terminal a $100 \times 0.5 \mathrm{~mm}$ copper strap is run un-interrupted to two 1/2" bolts welded to the hull as close to the equipment as possible

Wooden Ships:
From the transmitter earth terminal a $100 x 0.5 \mathrm{~mm}$ copper strap is run, preferably uninterrupted, to an earth bolt welded to an earth plate having a minimum area of 1 m 2 mounted under the water line. Should it, however, be necessary to break the copper strap, for example to pass through a deck, two 1/2" bolts should be used for this feed through. The copper strap should then be continued below deck, after connection to the same two bolts.

The copper strap must not be passed through iron pipes and should be kept a minimum distance of 0.5 m from iron parts of some extent. If this minimum distance cannot be kept the copper strap must be effectively connected to these parts using a strap having the same dimensions.

On wooden ships having a superstructure of metal, this superstructure should also be effectively connected to the copper strap.
5.2.2 Receiver Earth Terminal

The receiver earth terminal is located on terminal strip $A$ at the cabinet rear wall
3.5.2.1 Steel Ships

A flexible 2.5 MM2earth wire is run from the receiver earth terminal to a separate/2" earth bolt welded to the hull as close to the equipment as possible. This earth bolt must not be used for earthing other equipment, and the receiver earth wire shall be as far as possible from the transmitter copper strap.
3.5.2.2 Wooden Ships

A flexible 2.5 MM2 earth wire is run from the receiver earth terminal directly to the transmitter earth bolt on the earth plate. The earth wire should be run a minimum distance of 0.5 m from the transmitter copper strap.
3.5.3 Other Cables

Other cables should be placed as far away as possible from the transmitter and receiver earth leads and under no circumstances parallel with the transmitter copper strap closer than 0.7 meter and, for the receiver earth lead, closer than 0.2 meter.
3.5.4 Earthing the Battery

RF earth connections will cause neither battery nor mains leads to be connected to the hull. If it is desired to connect the battery to the hull, it is important to make the connection right at the battery, never in the transmitter. In cases where the installation is carried out so as to include the facility for charging during operation through a dropping resistor from a balanced ship's mains (110/220V DC), the battery must not be earthed.

Antennas
In order to minimise duplex noise, the transmitting and receiving antennas should be kept as far away from each other as possible. Stays, wires, steel masts etc. should either be earthed effectively or insulated.

Like wise in order to minimise duplex noise, every other electric installation such as cable braiding (screens) and instruments should be earthed effectively, and the instruments in question should be fitted with noise-interference suppression devices.

The Antennas should be suspended well in the clear, away from objects whose influence on the antennas may vary, such as derricks etc. Insulators should be of the best type having low leakage even when wet.
3.6.1 Transmitter Antenna Terminal:

The transmitter antenna terminal is located on the top of the cabinet.
3.6.2 Transmitter Antenna

The transmitter antenna tuning system will tune a minimum impedance of 150 p Fin series with 6 ohms at 1.6 MHz which normally corresponds to a wire length of approx. 14 m .

To ensure the greatest possible radiated power the transmitter antenna should be as long as possible and although a length of 14 meter can be tuned it is strongly recommended never to use lengths less than 20 m as the radiated power from short antennas will result in unsatisfactory communication. There are no limitations regardingthemaximumlengthoftheantranna. Theantennashouldbeterminatedin a feed-through insulator in the roof or side wall of the radio room. The feed-through insulator should be located in such a way that the distance between the insulator and the transmitter antenna terminal is as short as possible to avoid losses and radiated RF-power inside the radio-room which might disturb other equipment.

A short length of coaxial cable type RG8-U, of which only the braid and the outer insulation is used, is inserted between the feed-through insulator and the transmitter antenna terminal. Both ends of the coaxial screen are soldered to cable shoes of suitable dimensions for the feed-through insulator and the transmitter antenna terminal.
lf, for practical reasons, it should be necessary to mount the feedthrough insulator some distance $f$ rom the transmitter, the connection $f$ rom the insulatorto the vicinity of the transmitter should be done with a -length of copper tubing mounted on stand-off insulators. A length of coaxial cabie, as described above, should then be inserted between the last stand-off and the transmitter antenna terminal; any play between the transmitter and buikhead will then be taken up by the cable.
3.6.3 Receiver Antenna Terminal:

The receiver antenna terminal is located on term in al strip A at the cabinet rear wall.
3.6.4 Receiver Antenna:

Length: 7-50 m. The receiving antenna should be brought in by a length of coaxial cable, which should be as short as possible, especially in the case of a short antenna.

If a long coaxial cable is used in order to separate receiver and transmitter antennas it wilt often be advantageous to insert an impedance matching transformer at the antenna end of the coaxial cable.

While the receiver input impedance is always 50 ohm at frequencies above 4 MHz , the receiver is normally supplied with a high impedance input at frequencies below 4 MHz . It is however possible to change the receiver input impedance to 50 ohms also at frequencies below 4 MHz . Reference is made to the receiver instruction manual.

Antenna Relay:
As shown on the drawing on page 3-27, an antenna relay may be installed in the antenna circuit. The relay may be installed if the transmitting antenna is to be used for other purposes, for instance for an extra receiver, or if it is desired to perform the installation as a simplex installation with only one antenna. The relay coil should be rated for $24 V$ DC. It should be connected to terminals 3 (+) and 4 (-) on terminal strip A and protected with a suitable diode.
3.7 Remote Speaker

If a remote speaker is to be installed it should be connected to terminals 1 and 2 of terminal strip $A$

Note that the remote speaker will be connected to the receiver only when the SUPPLY switch is at RECEIVE ONLY and STAND-BY.

An audio power of 5 watts is available into a 4 ohms load. This power can be shared between several loudspeakers if so desired. The built-in speaker in the power pack has an impedance of 8 ohms. When connecting the remote speaker(s) the minimum value of the total impedance should be 4 ohms including the built-in speaker. If 5 watts is required in the remote speaker(s), the built-in speaker must be switched off.
3.8 Transmitter-On Indication

Indication of the transmitter-on condition can be obtained by means of a voltage ( 24 V at max. 0.2A) which is controlled by the transmit relay and can be taken off between terminals $3(+)$ and 4 (-) of terminal strip A.
3.9 Replacement of Power Pack

Two different power packs are available for operation of the equipment. The P 5000 is used for 24 V battery operation. If the equipment is to be operated from AC-mains the $P 5001$ power pack must be used. Replacing a power pack involves no modifications of transmitter or receiver.
3.10 Optional Terminal Strips

Three optional terminal strips are available. The terminal strips are supplied with cables .and plugs and are to be mounted on the cabinet back wall as shown on page 3-22.

Output Terminal Strip B gives a number of extra output facilities as shown on page $3-24$ and is to be connected to the Power Pack.

Input Terminal strip $C$, shown on page $3-25$, is to be connected to the Exciter.

Output Terminal Strip D, shown on page 3-26, is intended to be connected to the Transmitter Power Amplifier and can be used for example where it is desired to control an ATU. The corresponding socket in the T 5000 is only mounted and wired on special order, but a BAND switch wafer which can be wired to give the BAND position is always supplied as an integral part of the Transmitter Power Amplifier.

PROM-types which can be installed in the MEMORY of /238\}

| Manufacturer | Types |  |
| :---: | :---: | :---: |
| INTERSIL | IM 5600 |  |
| INTERSIL | IM 5610 |  |
| TI | SN 54188 | Only these PROM- |
| TI | SN 54188 A | types can be |
| TI | SN 54S188 | programmed by means |
| TI | SN 74188 | of the optional |
| TI | SN 74188 A | PROGRAMMING UNIT |
| TI | SN 74188 | $\underline{12411}$ |

Table 3.11
3.11.1 The building block of the memory, located on printed circuit board $/ 238 \$ , is the Programmable Read Only Memory, in short PROM.
A PROM is not re-programmable, because programming a bit position is like blowing a fuse.
3.11.2 On printed circuit board /238\} 2 4 PROM's can be mounted in separate sockets. The24soeketsareformed into eight columns (No. 1 to No.8), each containing three sockets (A, B, and C). The necessary information associated to a Frequency No is stored in three PROM's in a column. Each column can contain information of 32 different Frequency Nos. The association between the Frequency No's and the column No's is shown in table 3.12.9.
3.11.3 The programming of the $\mathrm{PROM}^{\prime} \mathrm{s}$ can done in two ways. The optional PROGRAMMING UNIT /241\ can be used for this job as described in paragraph 3.13. But only the eight peace's shown in the table 3.11 .1 can be programmed by means of printed circuit board $\underline{/ 241 \backslash}$

The other way is to let the local PROM-distributor do the programming. In this case all the PROM-type shown in table 3.11 .1 can be used in the MEMORY / 238\.
3.11.4 The MEMORY /238\ controls the Band Indicator of Transmitter Power Amplifier. This implies that the distribution of the coastal telephone frequencies in the bands $A$ to $H$ have to be decided in advance when programming the memory.

Each position of the Transmitter Power Amplifier BAND switch can, in principle, be adjusted to any frequency, but once adjusted (during installation on board a ship) the frequency coverage of each band is limited. In order to obtain a reasonable output power, the difference between the highest and the lowest frequency within a band should not exceed approx. 200 kHz.

Table 3.11.2 suggests a subdivision of the coastal telephone band based on article 7, section 4 (1976) of the Radio Regulations. If, however, special frequencies are to be covered, another subdivision may be necessary. Further, table 3.11 .2 shows the frequency ranges of the $H F$ bands covered by the Exciter.

| Transmitting <br> frequency (kHz) | BAND | Transmitting |  |
| :---: | :---: | ---: | ---: |
| $1605-1670$ | A | BAND |  |
| frequency (kHz) |  |  |  |
| $1950-2150$ | B | $4063-4219.4$ | 4 MHz |
| $2150-2350$ | C | $8195-8435.4$ | 8 MHz |
| $2350-2550$ | D | $12330-12652.3$ | 12 MHz |
| $2550-2750$ | E | $16460-16859.4$ | 16 MHz |
| $3000-3200$ | F | $22000-22310.5$ | 22 MHz |
| $3200-3400$ | G |  | 25 MHz |
| $3400-3600$ | H |  |  |

3.11 .5

Information about the transmitting mode of every Frequency No. in use must be stored in the memory, whether it is a radiotelephone frequency (A3A,A3J, orA3H)or a wireless telegraph frequency (Al, A2H, or Fl).
3.11 .6 Programming Frequency No. zero can be done in two ways. If it is convenient that this Frequency No. contains the information associated with an often used transmitting frequency, the programming instructions do not differ from the instructions concerning any other Frequency No.

On the other hand, if it is not desired to store any transmitting frequency information associated with Frequency No. zero, this can either be done as described in paragraph 3.12.14, if it is convenient to let the local PROM-distributor do the programming, or in paragraph 3.13.22 If the programming is performed by means of the optional PROGRAMMING UNIT /241\}
3.11.7 The easiest method of checking the programmed frequencies when the PROM's have been mounted in the MEMORY /238\ is by connecting a frequency counter to the output BNC socket, 231SK7, of the Exciter. The output socket is the one carrying no colour code. It is normal connected to the Transmitter Power Amplifier via a coaxial cable. Unplug the cable and connect the counter to this socket. The check is made with the SUPPLY switch in STAND BY. The frequency measured is the transmitting frequency, ft. Radiotelephone frequencies are measured in the A3H mode, DUPLEX. Wireless telegraph frequencies are measured in the Al mode, morse key down.

### 3.12 Ordering programmed $\mathrm{PROM}^{\prime s}$

3.12.1 To make it possible for the local PROM-distributor to do the programming the customer must fill in a Word Pattern Sheet, supplied by the distributor, for each PROM which is to be programmed.
3.12.2 First of all one must find out which of the 32 input addresses of the PROM corresponds to the wanted Frequency No. It is done in accordance with table 3.12.1.

| Frequency No. <br> (both <br> included) | Input Address <br> equal to: |
| :---: | :---: |
| $0-31$ |  |
| $32-63$ |  |
| $64-95$ |  |
| $96-127$ |  |
| $128-159$ |  |
| $160-191$ |  |
| $192-223$ | (Frequency No.) |
| $224-255$ | (Frequency No.) -32 |
| (Frequency No.) -64 |  |
| (Frequency No.) -128 |  |
| (Frequency No.) -160 |  |
| (Frequency No.) -192 |  |
| (Frequency No.) -224 |  |

3.12.3 Now the proper BAND corresponding to the transmitting frequency is to be found. If it is a coastal telephone frequency paragraph 3.11.4 describes what to do-

The encoding of the different BAND's is shown in table 3.12.2.

| BAND | Code |
| :---: | :---: |
| A | 0000 |
| B | 0001 |
| C | 0010 |
| D | 0011 |
| E | 0100 |
| F | 0101 |
| G | 0110 |
| H | 0111 |
| 4 MHz | 1000 |
| 6 MHz | 1001 |
| 8 MHz | 1010 |
| 12 MHz | 1011 |
| 16 MHz | 1100 |
| 22 MHz | 1101 |
| 25 MHz | 1110 |
| MF | 1111 |

Table 3.12.2
3.12.4 The transmitting mode is encoded as follows

| MODE |  |
| ---: | :---: |
| A3H, A3A, or A3J | 0 |
| Al, A2H, or Fl | 1 |
| $r \mid r ~ T a b l e ~ 3.12 .3 ~$ |  |

3.12.5. Calculate the synthesiser frequency fs from the transmitting frequency ft as shown in tabie3.12.4. Observe that the transmitting frequency ft is the carrier frequency in the modes A2H, Al, A3A, A3J and A3H. In the Fl-mode ft is the assigned frequency, provided that the centre frequency of the $A F$ output from the telex equipment is 1500 Hz . If the AF centre frequency is 1700 Hz , subtract 200 Hz from the assigned frequency to obtain ft.

| BAND | Fs <br> Synthesiser <br> frequency |
| ---: | :--- |
| $1605-4000 \mathrm{KHz}$ | $\mathrm{ft}+1400 \mathrm{KHz}$ |
| 4 MHz | $\mathrm{ft}+1400 \mathrm{KHz}$ |
| 6 MHz | $\mathrm{ft}-2800 \mathrm{KHz}$ |
| 8 MHz | $\mathrm{ft}-4200 \mathrm{KHz}$ |
| 12 MHz | $\mathrm{ft}-8400 \mathrm{KHz}$ |
| 16 MHz | $\mathrm{ft}-12600 \mathrm{KHz}$ |
| 22 MHz | $\mathrm{ft}-18200 \mathrm{KHz}$ |
| 25 MHz | $\mathrm{ft}-21000 \mathrm{KHz}$ |
| Table 3.12 .4 |  |

3.12.6 Each of the four least significant digits of the synthesiser frequency ("100 kHz", "10 kHz", "l kHz", and "100 Hz") are encoded as follows:

| Number | Code |
| :---: | :---: |
| 0 | 1001 |
| 1 | 1000 |
| 2 | 0111 |
| 3 | 0110 |
| 4 | 0101 |
| 5 | 0100 |
| 6 | 0011 |
| 7 | 0010 |
| 8 | 0001 |
| 9 | 0000 |
| Table 3.12.5 |  |

3.12.7 The most significant digit of the synthesiser frequency
("l MHz") is encoded as follows.

| Number | Code |
| :---: | :---: |
| 3 | 10 |
| 4 | 01 |
| 5 | 00 |
| Table 3.12 .6 |  |

3.12.8 In order to store this information 3 PROM's must be programmed at the same input address. These 3 PROM's are labelled A, B, and C. The total amount of information should now be arranged as follows:

| Input Address | $\begin{gathered} \text { Control } \\ \text { bit } \end{gathered}$ | Band | Mode | Synthesiser frequency |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 1 MHz | 100 KHz | 10 KHz | 1 KHz | 100 Hz |
| 1 | 07 | 06050403 | 02 | 0100 | 07060504 | 03020100 | 07060504 | 03020100 |
|  | PROM-C |  |  |  | PROM-B |  | PROM-A |  |

(Oo to 07 indicates outputs to be programmed).

Table 3.12.7
3.12.9 NB. The code of the "Control bit- is always to be 1 for each Frequency No. to be programmed.

Assume that the telex frequency 4170.0 KHz is to be programmed at Frequency No. 83.

The input address can be found from table 3.12.1:

Input address $=83-64=19$

According to paragraph 3.12.9 the Control bit is set to 1

According to table 3.12.2:

BAND $=4 \mathrm{MHz} \quad$ Code $=1000$

The code for the Fl mode is, according to table 3.12.3 :
1

The synthesiser frequency can be calculated from table 3.12.4
fs $=4170.0 \mathrm{KHz}+1400.0 \mathrm{KHz}=5570.0 \mathrm{KHz}$

By use of table 3.12 .5 and table 3.12 .6 the associated codes can be found :
Digit
1 MHz
100 KHz
10 KHz
1 KHz
100 Hz

| Number | Code |
| :---: | :---: |
| 5 | 00 |
| 5 | 0100 |
| 7 | 0010 |
| 0 | 1001 |
| 0 | 1001 |

The total amount of information is now to be arranged :

| (1) | (2) | (3) | (4) | (5) | (6) | (7) | (8) | (9) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input | Control | Band | Mode | Synthesiser frequency |  |  |  |  |
| Address | bit |  |  | 1 MHz | 100 KHz | 10 KHz | 1 KHz | 100 Hz |
| 19 | 1 | 1000 | 1 | 00 | 0100 | 0100 | 1001 | 1001 |
| 1 | 07 | 06050403 | 02 | 0100 | 07060504 | 03020100 | 07060504 | 03020100 |
|  | PROM-C |  |  |  | PROM-B |  | PROM-A |  |

Table 3.12.8a

Assume that a radio telephone frequency 3550.0 KHz is to be programmed at Frequency No. 228.

The input address can be found from table 3.12.1:

Input address $=228-224=\mathbf{4}$

According to paragraph 3.12.9 the Control bit is set to 1

According to table 3.11.2 and table 3.12.2:

BAND $=\mathrm{H}, \quad$ Code $=0111$

The code for the radio telephony mode is, according to table 3.12.3: 0

The synthesiser frequency can be calculated from tabie 3.12.4:
$\mathrm{fs}=3550.0 \mathrm{KHz}+1400.0 \mathrm{KHz}=\mathbf{4 9 5 0 . 0 \mathrm { KHz }}$

By use of table 3.12.5 and table 3.12.6 the associated codes can be found:

| Digit | Number | Code |
| ---: | :---: | :---: |
| 1 MHz | $\mathbf{4}$ | 01 |
| 100 KHz | 9 | 0000 |
| 10 KHz | 5 | 0010 |
| 1 KHz | 0 | 1001 |
| 100 Hz | 0 | 1001 |

The total amount of information is now to be arranged:

| (1) | (2) | (3) | (4) | (5) | (6) | (7) | (8) | (9) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input | Control | Band | Mode |  | Synthes | ser fr | quency |  |
| Address | bit |  |  | 1 MHz | 100 KHz | 10 KHz | 1 KHz | 100 Hz |
| 4 | 1 | 0111 | 0 | 01 | 0000 | 0100 | 1001 | 1001 |
| 1 | 07 | 06050403 | 02 | 0100 | 07060504 | 03020100 | 07060504 | 03020100 |
|  | PROM-C |  |  |  | PROM-B |  | PROM-A |  |

Table 3.12.8b
3.12.11 When the three $\mathrm{PROM}^{\prime}$ s are to be installed in the MEMORY /238\ make sure that they are mounted in the correct column:

| Column <br> No. | Frequency No. <br> (both included) |
| :---: | :---: |
| 1 | $0-31$ |
| 2 | $32-63$ |
| 3 | $64-95$ |
| 4 | $96-127$ |
| 5 | $128-159$ |
| 6 | $160-191$ |
| 7 | $192-223$ |
| 8 | $224-255$ |

Table 3.12.9
3.12.12 Also take care that $P R O M-A$ is mounted in the socket marked with an $A$, PROM-B in socket $B$, and $P R O M-C$; $n$ socket $C$.

The top mark of the PROM-package is to be directed as shown on printed circuit board /238\}
3.12.13 A self adhesive sticker should be placed on the package of each PROM, indicating in which column and socket ( $A, B$, or $C$ ) it is to be mounted.
3.12.14 Concerning the programming of Frequency No. zero some considerations are to be made; refer to paragraph 3.11.6.

If no transmitting frequency information is to be stored at Frequency No. zero, the following information must be programmed at this
Frequency No.:
Input address $=0$
Control bit $=1$
BAND $=1111$ (Keying the transmitter will be inhibited)
MODE
$=0$
Synth. freq. $=5999.9 \mathrm{KHz}$
Or in the arranged form:

| (1) | (2) | (3) | (4) | (5) | (6) | (7) | (8) | (9) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input | Control | Band | Mode | Synthesiser frequency |  |  |  |  |
| Address | bit |  |  | 1 MHz | 100 KHz | 10 KHz | 1 KHz | 100 Hz |
| 0 | 1 | 1111 | 0 | 00 | 0000 | 0000 | 0000 | 0000 |
| 1 | 07 | 06050403 | 02 | 0100 | 07060504 | O3020100 | 07060504 | 03020100 |
|  | PROM-C |  |  |  | PROM-B |  | PROM-A |  |

3.13.1 Three sockets are located in the PROGRAMMINGUNIT. From MEMORY /238\ the three $P R O M^{\prime} s$ in the column No. found from table 3.12 .9 are moved to board /241 and mounted in these three sockets.

Take care that the one placed in socket $A$ on /238\} is mounted in the socket marked with an $A$, socket $B$ in socket $B$, and socket $C$ in socket C. Also take care that the top mark of the PROM package is directed as shown on /2411.
3.13.2 A minor dis-advantage associated with the use of the PROM as the memory building block is that a few per cent of the PROM's cannot be programmed in one or more bit positions due to tolerance problems in the manufacturing process. In this case section 3.13 .17 describes what to do.
3.13.3 Because PROM is not re-programmable the greatest care should be taken concerning the programming procedure.
3.13.4 For each Frequency No. the following information must be stored in the PROMS. This information should be written on a Programming Work Sheet as shown on page 3-34.
3.13.5 The proper BAND corresponding to the transmitting frequency is to be found from table 3.11 .2 if it is a coastal telephone frequency, otherwise it is self explanatory.
3.13.6 It must be decided whether it is to be a radiotelephone frequency (RT=A3H,A3A, or A3J) or a wireless telegraph frequency (WT = Al, A2H, or Fl).
3.13.7 Furthermore the synthesiser frequency must be calculated from the transmitting frequency as shown on the screening lid to the right of the PROGRAMMING UNIT or in table 3.12.4.
3.13.8 By means of the ON-OFF switch on $/ 241$ this unit is turned on.
3.13.9 A the sliders of the programming tool are now adjusted according tot his information. In the other window some holes will appear and these are the positions which must be programmed.
3.13.10 By means of the Keyboard on the front panel the Frequency No., which is to be programmed, is selected. (The displays may flash during the programming procedure, but this is unimportant).
3.13.11 If the lamp of the grey pushbutton is now lit, it means that the Frequency No. chosen has already been programmed once, and further programming is automatically inhibited.
3.13.12 On the other hand if the above mentioned lamp is unlit, it means that programming can now be done.
3.13.13 The Programming Pin is now placed in the extreme right hole in the window. The lamp of the red pushbutton will light as long as the Programming Pin does not have proper contact with the Underlying socket-terminal or if this hole position has already been programmed.
3.13.14 Now activate the red pushbutton.
3.13.15 If the programming was done successfully the lamp of the red push button will now light and the Pin is moved to the next hole.
3.13.16 If the programming was not done successfully, the red push button must be activated repeatedly until the lamp is lit.
3.13.17 If it turns out that it is impossible to program this hole position, the only thing to do is to choose another Frequency No. and start to program all over again. Later on, when the PROM's are installed on /238\ , the transmitter is automatically prevented from being keyed, if this unsuccessfully programmed Frequency No. is chosen; this will also be indicated by the displays, which will begin to flash.
3.13.18 If all the hole positions in the window have been successfully programmed, the Programming Pin must be placed in the hole in the middle of the tool, and this position is then to be programmed.
3.13.19 This will make the lamp of the grey pushbutton light, thus indicating that this Frequency No. is correctly programmed; any attempt to program further information at this Frequency No. is now automatically prevented.
3.13.20 If, later on, it turns out that nevertheless a mistake was made, when the sliders were adjusted, and the »Enable» hole position had been programmed, it is still possible to »Disable» the Frequency No. This is done as follows.
3.13.21 Adjust the "MHz"-slider so that a black field appears in the window; two holes will now appear in the other window; now push and keep down the grey pushbutton until both of these two hole positions are programmed, as normal by means of the Programming Pin and the red pushbutton. When this has been accomplished, it will prevent the transmitter from being keyed when the PROM's are mounted on /238\. If this Frequency No. is chosen; this will also be indicated by means of the flashing displays.
3.13.22 Concerning the programming of Frequency No. Zero some considerations are to be made, refer to paragraph 3.11.6.
3.13.23 If no transmitting frequency information is to be stored at Frequency No. zero, the following procedure must be followed.

Adjust the Band-slider so that an "MF" appears in the window; only the associated four hole positions and after wards the Enable position in the middle of the tool must be programmed.

This will, when the three $\mathrm{PROM}^{\prime}$ s are mounted on /238<br>, prevent the displays and the Band Indicator from flashing and also prevent the transmitter from being keyed when Frequency No. zero is chosen.
3.13.24 When the programming of the Frequency No's has been accomplished the PRO- GRAMMING UNIT is turned off, the Programming Pin is placed in its holding clips, and the three $\mathrm{PROM}^{\prime}$ s are moved back to the MEMORY /238\.

Take care that the $P R O M^{\prime}$ s are mounted in the correct way and in the correct sockets.
3.13.25 A self adhesive sticker should be placed on the package of each PROM indicating in which column and socket ( $A, B$, or $C$ ) it is to be mounted.
3.13.26 Programming procedure step by step
l. Calculate for each Frequency No. the synthesiser frequency from the transmitting frequency as shown in table 13.12.4, find the correct BAND (for a coastal telephone frequency table 3.11 .2 is used), and decide whether it is to be a radiotelephone frequency ( RT ) or a radiotelegraph frequency (WT)
2. Mount the three PROMS in their respective sockets on /241\. (Take care that they are correctly positioned).
3. Turn on the PROGRAMMING UNIT
4. Adjust the seven sliders according to step 1
5. Select by means of the Keyboard the Frequency No. to be programmed (The displays may now flash, but this is unimportant).
If the lamp of the grey pushbutton lights, the Frequency No. chosen has already been programmed and a new Frequency No. must be selected.
6. Place the Programming Pin in one of the holes in the window of the tool; beginning from the right.

If this hole position is un-programmed the lamp of the red pushbutton will extinguish.
7. Activate the red pushbutton

The lamp of the red pushbutton will now light if the programming was successful.
As long as there are un-programmed hole positions the steps 6 and 7 are repeated.
If the lamp of the red pushbutton does not light, step 7 is repeated until it lights. (]f this turns out to be impossible then proceed to step g).
8. Program "Enable" hole position in the middle of the Programming Tool, thus enabling this Frequency No.

If there are more Frequency $N N^{\prime}$ 's to program then go back to step 4, otherwise to step 16.

If there is a hole position, which cannot be programmed
9. Select a new un-programmed Frequency No. and start from step 6

If it turns out, after having performed step 8, that a mistake was made concerning the information already programmed:
10. Adjust the "MHz"-slider, so that a black field appears in the window
11. Place the Programming Pin in one of the two associated holes
12. Push and keep down during step 13 the grey pushbutton.
13. Activate the red pushbutton.
14. Move the Programming Pin to the other hole and repeat the steps 12 and 13.
15. Go back to step 4 and repeat through to step 8
16. Turn off the Programming Unit.
17. Place the Programming Pin in its holding clips.
18. Place a self adhesive sticker on the package of each PROM, indicating in which column and socket (A, B, or C) it is to be mounted on MEMORY /238
19. Move the three $\mathrm{PROM}^{\prime}$ s to /238\. (Take care that they are mounted in the correct way and in the correct sockets).
3.14.1 The drum switch of the antenna tuning circuit becomes accessible by pulling the Transmitter Power Amplifier unit partly out of the cabinet. As connections to earth and antenna are maintained, it is possible to perform the adjustments with the Transmitter Power Amplifier in this position. When adjustment has been completed it ought to be checked with the unit in place. Minor corrections may be necessary.

The safety switch, that normal prevents keying when the transmitter power amplifier unit is pulled out, has to be disabled. The piston of the switch has to be pulled outwards to the locked position.

Note: Extreme care has to be taken as high tension is now accessible when the SUPPLY switch is in position TRANSMIT.

The various components of the antenna tuning circuit are selected by means of the drum switch which can be programmed individually for each band by inserting programming tabs into the appropriate wafers. The distribution of frequencies into the bands has been determined when programming the MEMORY /238\ as described in section 13.11.

It is important that the adjustments are made with the antenna in its normal working positioned that no cranes, derriks, etc. are near, as these will influence the tuning.

The meter on the front panel indicates the actual antenna current. Since the antenna impedance varies considerably over the frequency range, the antenna current will vary correspondingly, even though the output power is the same. If the antenna length is approximately half a wavelength or a multiple thereof, the meter reading will be very low. If this is un-desirable, a higher reading can only be obtained by shortening or lengthening the antenna.

Due to the wide adjustment range when programming the drum switch, attention must be drawn to the possibility of erroneously adjusting to the second harmonic frequency. The only sure method of checking this is by means of a frequency counter or an oscilloscope connected to a loop around the antenna insulator of the transmitter. Check one frequency in each band. (A3H or A3A DUPLEX, TUNE button not pressed).

The configuration is basically an L-network, the series inductance of which consist of may consist of variometer -A-. The rotor and stator of variometer can be used on the lower frequencies, giving a high value of inductance. The parallel connection is to be used on the higher frequencies, giving a lower value of inductance. The transition frequency depends on the antenna.

The shunt capacitance, in parallel to the $P . A . v a l v e s, ~ c a n ~ b e ~ s e l e c t e d ~$ between 6 and 1600 pF in steps of 6 pF by means of 1 to 9. Increasing the shunt capacitance while still having resonance (by adjusting the variometer by means of the tuning knob) decreases the voltage swing on the P.A. valves ( the load presented to the valves becomes lower) and vice versa.

If the antenna is long, it may be necessary to use the shortening capacitors or the output capacitor (wafers 22, 23, and 24 or any combination of these) in order to obtain resonance. If the antenna is short, it may be necessary to use the output capacitor (wafer 24) on the lowest frequencies.

The configuration is basically, a 27 -network. The series inductance may consist of either variometer -A- or variometer -B-. The rotor and stator of variometer -B- may be connected either in series or in parallel. The parallel connection is to be used only in the higher frequency bands. The transition frequency depends on the antenna.

The shunt capacitance, in parallel to the P. A. valves, is pre-set on each band by means of wafers 13, 14, end 15.

The output capacitance can be selected between 6 and 1600 pF by means of wafers 1 to 9. increasing the output capacitance, while still having resonance, increases the voltage swing on the $P$. A. valves and vice versa.

On frequencies above 8 MHz the output capacitance can be reduced, it necessary, by means of the coils at wafers 24 and 25 . The coil at after 24 is to be used only on the 22 MHz and 25 MHz band's, where the impedance of the series connected capacitor, is negligible.
Due to un-avoidable stray capacitance's, resonance in variometer -Acan occur on one or more of the short wave band's. It shows up as a sudden dip in the ANTENNA CURRENT meter-reading when the TUNING control is rotated slowly over its range. Such resonance's within a band must be avoided as they can cause flash-over. This is avoided by showing variometer -A- of the band in question by inserting tabs in wafers 27 or 17 and 19 or all three.
3.14.4 Adjustment Procedure:
3.14.4.1 1. Set SUPPLY switch to STAND BY and POWER switch to MEDIUM POWER SIMPLEX.
2. Insert tabs for the band in question as shown on Pages 3-28 to 3-32.
3. Set SUPPLY switch to TRANSMIT and press TUNE button while rotating the TUMNG control until resonance is obtained, i.e. maximum deflection on ANTENNA CURRENT meter without being at the extreme of the tuning range (inspect rotor position of the variometer used).
4. Set CHECK SWITCH to LEVEL and pull the knob while pressing the TUNE button. Observe Level-reading on meter. Reading should be as near 3 as possible but never above.
5. Set SUPPLY switch to STAND BY and adjust Level by selecting appropriate tab settings on wafers 1 to 9 (refer to sections 3.14.2 and 3.14.3). Revert to point 3.
3.14.4.2 If resonance is obtained at, or is outside, one of the extreme positions of the TUNING knob, select another variometer configuration.
3.14.4.3 If resonance cannot be obtained, remove tab on wafer 21 and insert one or both of the shortening capacitors (wafers 22 and 23), always use the largest possible capacitance.
3.14.4.4 If on bands $A$ to $H$ resonance still cannot be c>btained, insert the output capacitor (wafer 24), if necessary in combination with one or both of the shortening capacitors.
3.14.4.5 If Level-reading is too high, even with none of the capacitors 1 to 9 connected, insert the coil at wafer 25 on the 8,12 or 16 MHz bands, or the coil at wafer 24 on the 22 or 25 MHz bands.
3.14.4.6 If flash-over occurs on 22 MHz at contact 25 , insert tab in wafer 25.
3.14.5 $2182 \mathrm{KHz}:$

The adjustment is performed as described above for a frequency in the lower CT band. In order to reduce the range of the TUNING control proceed as follows:

1. Move tab on wafer 19 to wafer 18.
2. Place one end of the clip-on lead on the middle of $>2182 \mathrm{KHz}$ Fine Tuning Coil«.
3. With the other end of the clip-on lead, select that winding on the stator of Variometer »A« or the »2182 kHz Coil«, below, which gives the highest level reading.
4. Rotate TUNING control to it se enter position and select that winding on the" 2182 kHz Fine Tuning Coil" which gives maximum Level-reading.

absorbers can be reversed and thereafter only one hole per shock absorber need be drilled.

Horizontal drilling plan for mouting


Dementions are in mm
Tollerances : + / - 1 mm

| unit : |  | Appr. Veight: |
| :---: | :---: | :---: |
| TRP 5000 | DC | 83.5 kg |
| TRP 5000 | AC | 89.5 kg |




TRP 5000 CABINET, ALL OPTIONAL TERMINALS STRIPS INSTALLED

| b | Blue |
| :---: | :--- |
| be | Beige |
| bk | Black |
| bn | Brown |
| g | Green |
| or | Orange |
| P | Pink |
| r | Red |
| S | Slate (grey) |
| v | Violet |
| w | White |
| Y | Yellow |
| tt | transparent |



Note 1:

| Max cable length <br> to battery | Minimum <br> conductor size |
| :---: | :---: |
| 5 Meter | $2 \times 10 \mathrm{~mm} 2$ |
| 9 meter | $2 \times 16 \mathrm{~mm} 2$ |
| 13 meter | $2 \times 25 \mathrm{~mm} 2$ |

NOTE 2
AN AUDIO POWER OF 5 WATTS IS AVAILABLE INTO A, 40 OHMS LOAD.THIS POWER CAN BE SHARED BETWEEN SEVERAL LOUDSPEAKERS IF SO DESIRED. THE BUILT-IN SPEAKER IN THE POWER PACK HAS AN IMPEDANCE OF 80 OHMS. WHEN CONNECTING, REMOTE SPEAKERS THE MINIMUM VALUE OF THE TOTAL IMPEDANCE SHOULD BE MORE THAN 4OHMS INCLUDING THE BUILT-IN SPEAKER IN ORDER TO OBTAIN MAXIMUM POWER OUTPUT. IF 5 WATTS IS REQUIRED IN REMOTE SPEAKER(S) THE BUILT-IN SPEAKER MUST BE DISCONNECTED.


## INPUT TERMINAL STRIP C OPTIONAI

(Terminal strip to be mounted on cabinet back wall)


# T 5000 <br> OUTPUT TERMINAL STRIP D OPTIONAL 

(terminal strip to be mounted on cabinet back wall)
To be used for example to control an ATU. Note that the corresponding socket SK2 in T5000 only is mounted and wired on special order.




> INITIAL TAB-SETTINGS ON BAND-SWITCH DRUM LOWER CT-BAND (below approx. 3 MHz )


INITIAL TAB-SETTINGS ON BAND-SWITCH DRUM UPPER CT-BAND (above approx. 3 MHz )


INITIAL TAB-SETTINGS ON BAND-SWITCH DRUM
$4-6 M H Z ~ B A N D S ~$


INITIAL TAB-SETTINGS ON BAND-SWITCH DRUM $8-16 \mathrm{MHz}$ BANDS


INITIAL TAB-SETTINGS ON BAND-SWITCH DRUM $22-25 \mathrm{MHz}$ BANDS

Ship : $\qquad$ Harbour: $\qquad$
Technician : $\square$ Date: $\qquad$ T 5000 serial no.:
Supply voltage : $\qquad$ V DC/AC-50/60Hz Antenna: $\qquad$

CHECK SWITCH readings are taken with POWER switch in position FULL POWER DUPLEX and MODE switch in position A3J

| CHECK <br> SWITCH | TUNE <br> button not <br> pressed | TUNE button <br> pressed |
| :--- | :--- | :--- |
| VA |  |  |
| VS1 |  |  |
| VS2 |  |  |
| VG |  |  |
| I1 |  |  |
| I2 |  |  |
| VD |  |  |
| Freq. KHz (not 2182) $:$ |  |  |

Level readings are taken with POWER switch in position MEDIUM POWER SIMPLEX and Tune button pressed. ANTANNA CURRENT readings are taken with POWER switch in position FULL POWER SIMPLEX and TUNE button pressed. Readings are taken on that frequency in each band wich gives the highes Level reading.
Mark in plan below positions an drum switch in witch programming tabs have beenn inserted.

| Band | Freq. KHz | LEVEL | Ant. Current | $r$ | $\sim$ | m | - | $\llcorner$ | $\bigcirc$ | $\checkmark$ | $\infty$ | の | $\bigcirc$ | $\begin{aligned} & -1 \\ & -1 \end{aligned}$ | $\underset{\sim}{\sim}$ | $\underset{\sim}{m}$ | $\stackrel{\sim}{\sim}$ | $\begin{gathered} 6 \\ r-1 \end{gathered}$ | $\stackrel{\lambda}{\lambda}$ | $\stackrel{\infty}{\infty}$ | $\left\|\begin{array}{l} a \\ r \end{array}\right\|$ | $\stackrel{\rightharpoonup}{\mathrm{v}}$ | $\stackrel{-}{\sim}$ | $\underset{N}{N}$ | $\stackrel{m}{N}$ | ~ | $\left\|\begin{array}{l} \stackrel{e}{N} \\ \mathbf{n} \end{array}\right\|$ | $\stackrel{\wedge}{\sim}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2182 | ---- |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| A |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| B |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| C |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| D |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| E |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| F |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| G |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| H |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 4 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 6 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 8 Mhz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 12 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 16 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 22 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| 25 MHz |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |


| Transmitting <br> Frequency $(\mathrm{KHz})$ | Frequency <br> Correction (KHz) | Freq. <br> No. | Band | Mode | Synthesizer <br> Frequency (KHz) |
| :---: | :---: | :---: | :---: | :---: | :---: |
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## 4. Technical Data

## Modes of operation

Duplex, semi-duplex and simplex A3A, A3H, A3J, and Fl.
Semi-duplex and simplex Al and A2H.
At 2182 kHz A3H simplex only

## Power Output

Al, A2H, A3A, and A3J: 400 watts Peak Envelope Power
Fl: 250 watts
with reduction to less than 20 watts.

## Operating frequencies

2182 kHz plus up to 256 synthesised frequencies. The frequencies can be freely distributed in the maritime mobile bands up to 27.5 MHz .

Frequency Accuracy Better than $\pm 40 \mathrm{~Hz}$

## Frequency Stability

Better than $\pm 20 \mathrm{~Hz}$ in any 15 minute period.

## Modulation characteristic

Within 6 dB from 350 Hz to 2700 Hz

## Alarm Generator

A two-tone alarm generator is incorporated.

## Supply Voltage

24 V battery with P5000 Power Pack or 110/115/120 or $220 / 230 / 240$ V single, or two phase AC, 50-60 Hz with P5001 Power Pack.

Supply Voltage Variations
DC: - 10 to + 30\%
$\mathrm{AC}: \pm 10 \%$

## Consumption

```
RECEIVE ONLY
```

STAND BY
TRANSMIT (un-keyed)
A3J (un-modulated)
A3H (un-modulated)
TRANSMIT ALARM
Fl
Al (50\% duty cycle)

| 24 V battery | AC mains |
| :---: | ---: |
| 2.5 A | 85 VA |
| 6.5 A | 235 VA |
| 9 | A |
| 27 | A |
| 34.5 | 270 A |
| 39.5 A | 800 VA |
| 36.5 A | 1200 VA |
| 35 | A |

## Dimensions and weight

| Height (cabinet only): 822 mm |  |
| :--- | :--- |
| With (cabinet only): | 533 mm |
| Depth (cabinet only): | 334 mm |
| Weight (DC): | 83.5 kgs |
| Weight (AC): | 89.5 kgs |

## 5. Technical Description

### 5.1 Mechanical

5.1.1 The equipment consists of four main units: Transmitter Power Amplifier, Power Pack, Exciter and Receiver. The Transmitter Power Amplifier is built on an alodine treated aluminium chassis. The other units are built on zinc plated and passivated iron chassis.
5.1.2 The cooling of the two P.A. valves is by forced air, provided by a blower installed at the back of the Transmitter Power Amplifier chassis. The air-intake is on the front panel where an air filter is mounted. The filter element can be removed for cleaning.
5.1.3 The transmitter power amplifier T5000 contains four printed circuit boards. The boards $/ 251 \backslash$ and $/ 254 \backslash$ become accessible after removal of the lid of the tube compartment at the back of the chassis./252\ is mounted on the CHECKSWITCH and $/ 253 \backslash$ becomes accessible after removal of the front plate.
5.1.4 The Exciter E5000 contains five plug-inboards /234\ to /238\ and two boards in separate screened cans /232\ and /238\ that become accessible after removal of the respective lids. Board /239 front panel. This board and the motherboard /231\ become accessible when the front panel is tilted forward, this is possible after removal of the two upper screws in each side fixing the front panel.
5.1.5 The 24 V DC power pack P 5000 contains four printed circuit boards $/ 261 \backslash$ to /264 . The boards /261\} \text { and /262\} are mounted below the chassis, while board /263\} \text { and /264\} are mounted on the back of the chassis front plate.
5.1.6 The AC power pack $P 5001$ contains three printed circuit boards /266\ to /268\. The boards /266\ and /267\ are mounted below the chassis, while board /268\} is mounted on the back of the chassis front plate.

### 5.2 Circuit Description, General

Each printed circuit-board and also the chassis-mounted components, have been allocated an identification number:

Exciter, E5000, /230\ to /241\
Transmitter Power Amplifier, T5000 /250\ to /254\}
Power Pack, P5001 /260\ to /264
Power Pack, P5001 /265\ to /268\}

The designation of a component or terminal includes this number as a prefix, e.g. 232R3 (resistor R3 on board /232<br>), or 232-12 (terminal No. 12 on board /232\.

For convenience in this section and on the circuit diagrams, the prefix is omitted except where there is a risk of ambiguity.

### 5.3 Circuit Summary, Exciter E 5000

5.3.1 The circuit diagram is divided into a wiring diagram on page 8-67 showing the Motherboard $/ 231$ and the interconnections between the printed circuit boards, of which the Exciter is composed, and circuit diagrams of the individual circuit boards. The block diagram on page 860 illustrates the operation of the Exciter.
5.3.2 The signal path is contained on boards $/ 232 \backslash$ and /232\. The AF input signal, after having passed an input selector and a compressor, is converted to a $1,4 \mathrm{MHz}$ (or $1,973 \mathrm{MHz}$ ) double side-band signal by mixing with a 1.4 MHz (or $1,973 \mathrm{MHz}$ ) signal derived from the master oscillator. The upper side-band is removed in a crystal filter and the lower sideband is applied to an amplifier, the gain of which is set in accordance with the mode selected. At the output a 1.4 MHz signal of appropriate level for carrier re-insertion is applied in the modes A3A and A3H.
5.3.3 The 1.4 MHz (or $1,973 \mathrm{MHz}$ ) single side-band signal is fed to the RF Translator $/ 233 \backslash$ that converts it to the desired transmission frequency. The lst mixer converts the signal to an upper side-band signal in the range 1.6 to 4.3 MHz , a low-pass filter removes the image frequency. The gain of the amplifier following the filter can be controlled in steps from the POWER-switch. The level setting circuitry at the output of the amplifier allows the level to be set independently at each band.

The 2nd. mixer converts the signal to the different bands by mixing it with fixed injection frequencies. The mixer is followed by switchable band pass filters and an amplifier.
5.3.4 In the Al,A2 Hand TRANSMIT ALARM modes the AF signal is supplied from the tone generators on board $/ 234$. The 1.5 kHz oscillator is used for generating the carrier frequencies in the modes Al and A2H. A 1.5 kHz shift command from the mode control circuit makes the synthesiser decrease its output frequency by 1.5 kHz in these modes which means that the output frequency of the exciter be comes the correct transmission frequency. The 2.2 kHz oscillator in addition is used for generating the side-band intheA2H mode giving a modulation frequency of 700 Hz .

The Key-line output from the keying circuit to the Power Pack tells the HT converter to start when the line goes high. However, this is inhibited until the phase locked loops of the synthesiser and the Harmonic Generator have locked or if an invalid frequency No. has been selected. If the 1.4 MHz Reference Signal is missing, the key-line is also inhibited, as the synthesiser cannot lock.
5.3.5 The Frequency Synthesiser, contained on board /236\ and /237<br>, consists of a single loop using the so-called Side-step-technique. The loop is able to generate any frequency from $3,000.0 \mathrm{KHz}$ to $5,999.9 \mathrm{KHz}$ in steps of 100 Hz . In the Fl, Al, A2H modes the output frequency is automatically decreased by 1.5 KHz .

The Frequency Synthesiser loop is locked to a l KHz reference signal derived from the 1.4 MHz reference frequency, so that the output frequency will exhibit exactly the same stability as the master oscillator. The output frequency of the loop is controlled from the Memory /238\.

The Side-step principle is performed as follows. Assume that the output frequency of the Frequency Synthesiser is (a.b.c.d.,e) KHz. Th is frequency is divided by (abcd + 1)e times and (abcd) * (10-e) times, so that the division ratio on the average is: ( (abcd + 1)* e + (abcd)* (10e)) = abed.e. Thus the average output frequency of the variable counter chain will become equal to the 1 KHz reference frequency of the loop, as it should, if the loop is properly locked.
5.3.6 The injection frequency to the 2 nd. mixer of the RF Translator is supplied by the Harmonic Generator /235\. The frequencies are all harmonies of the 1.4 MHz (or $1,973 \mathrm{MHz}$ ) reference frequency. The desired harmonic is selected by a Band Selector control- led from the Memory. The Band Selector also selects the correct band-pass filter and level setting potentiometer of the RF Translator.

The Memory /238\ in which the necessary information associated to each frequency No. is stored controls the synthesiser and the Harmonic Generator, and supplies information to the Band indicator of the Transmitter Power Amplifier. The memory address is selected from board /239\ which contains the keyboard and a display showing the selected Frequency No.

The $1,4 \mathrm{MHz}$ (or $1,973 \mathrm{MHz}$ ) reference frequency is normally supplied from the master oscillator of the receiver. However, if the receive $r$ is not installed in the cabinet, Master Oscillator /240\ (or /240a\ is available for mounting in the Exciter.

### 5.4 Circuit Description, Exciter E 5000 or E 5001

5.4.1 /232\ 1.4 MHz (or 1,973 MHz) EXCITER

The AF input signals are connected to the compressor through an input selector. The microphone and line inputs are open only if the terminals 4 and 8 are both HIGH. The telex $A F$ input is open only if terminal 8 is LOW. The tone input is always open.

The gain of the compressor is controlled by means of the field effect transistor TR2, which functions as a variable emitter resistor for the left hand transistor in IC1. The control voltage is provided by a rectifier consisting of the pair of ICI transistors to the right, which detect the side-band level at the output of the crystal filter X1. When terminal 4 is LOW the compressor is off, as the resistance of TR2 is kept at its maximum value.

The compressed audio signal and a 1.4 MHz (or $1,973 \mathrm{MHz}$ ) signal from the carrier level regulator IC2 are fed to the balanced mixer IC3. The output is a 1.4 MHz (or $1,973 \mathrm{MHz}$ ) double side-band suppressed carrier signal, which is amplified in TR4 and fed through crystal filter X1 that removes the upper side-band and suppresses the carrier still further.

The gain of the amplifier stage following the filter is controlled from terminals 13, 14 , and 15 by inserting different emitter resistors. In the A3H-rnode all terminals are $H I G H$ and the gain is determined by R70. Carrier re-insertion is performed by applying the 1.4 MHz signal from the carrier level regulator to TR6 via an attenuator controlled from the same terminals.

The lst. mixer converts the 1.4 MHz (or $1,973 \mathrm{MHz}$ for LSB) LSB-signal to an USB-signal in the frequency range 1.6 to 4.3 MHz , by mixing it with the injection frequency from the VCO of the synthesiser. The image signal is removed in the 4.3 MHz low-pass filter and the wanted signal is applied to TRI. The gain of this stage is controlled from the POWER switch via terminals 1 and 2. The load-resistance at the collector is adjustable and can be set individually at each band by means of variable resistors.

The 2nd. mixer converts the signal to the transmission frequency. The injection frequency is fixed for each band. In the bands 1.6 to 4 MHz and the 4 MHz HF band the injection frequency is zero, which means that the output frequency is the same as the input frequency. The mixer works as an amplifier since the one pair of transistors are cut off by means of Dl 1 or 012.

The mixer is followed by a number of band pass-filters and an amplifier. A filter is selected by taking one of the terminals 12 to 21 LOW, while keeping the others HIGH. At the same time the corresponding level setting potentiometer is selected.

### 5.4.3 /234 CONTROL CIRCUIT AND TONE GENERATORS

The mode control signals from the MODE-switch, and the 2182 khz and MF information from the transmitter power amplifier, are the input signals to the programmable Read Only Memory (PROM), IC5, of the Mode Selector. The PROM has been programmed to give at the output the desired control signals corresponding with the wanted mode.

The Mode Selector controls the input selector and the mode setting of the 1.4 MHz (or $1,973 \mathrm{MHz}$ ) Exciter $/ 232$ via the respective control leads. The 1.5 kHz SHIFT output. tells, when HIGH, the synthesiser to decrease its frequency by 1.5 kHz .
In the Fl-mode, the anode voltage of the P.A. valves is lowered by means of are lay in the Power Pack controlled from transistor TR5. Automatic selection of FULL POWER is carried out by means of transistor TR4, when 2182 kHz (terminal 14a LOW) and TRANSMIT ALARM (output 06 of IC5 LOW) are selected. The mode selector further controls the Key Selector.

Keying of the transmitter is only possible from the MORSE KEY input in the A2H and Al modes, from the TELEX KEY input in the F1 mode, from the HANDSET KEY input in the A3A, A3J, A3 Hand TRANSMIT ALARM modes, and from the DUPLEX input in the Fl, A3A, A3J, A3H or TRANSMIT ALARM modes, provided 2182 kHz is not selected, as the DUPLEX input is then inhibited. Furthermore the information on terminal $24 a$ (WT) from MEMORY /238\ has to be in accordance with the mode selected (except on 2182 $\mathrm{kHz}): \mathrm{HIGH}$ at A2H, Al, and Fl, LOW at A3A, A3J, A3H, or TRANSMIT ALARM. The keying signal at IC2, pin 8 controls, via an inverter, transistor TR9, supplying +12V to RF TRANSLATOR /233\ when keyed, and activating the key-line to the power pack. In SIMPLEX it supplies base current to TR8, controlling the muting of the receiver.

The Two-tone Alarm Signal Generator incorporates the 45 sec. Timer, the 2 Hz Astable Multivibrator, and the 2.2 kHz and 1.3 kHz Oscillators. The 45 sec. Timer is enabled from the Mode Selector in

TEST ALARM and TRANSMIT ALARM, and can be started by pushing the ALARM START push-button that applies +12 V to terminal 2c. Via the voltage divider R32 and R33, a keying signal is applied to the Key Selector. The 2 Hz Astable Muitivibrator starts and supplies base current to TR6 and TR11 alternatively. The audio signal from the oscillators is, via the Tone Keyer, applied to the 1.4 MHz EXCITER /232\} and, via the Side Tone Keyer, applied to the receiver audio amplifier.

In the Al-mode the 1.5 kHz oscillator is started. The audio signal is keyed in the Side-tone Keyer and the Tone Keyer which are both controlled from the morse key. Capacitor C 8 and adjoining components at the gain control input of $I C 7$ serve to shape the tone pulses correctly.

In the A2H-mode, the 2.2 kHz oscillator as well as the 1.5 khz oscillator are on. R54 is connected to ground in IC4, thereby reducing the peak level of the audio signal at the Tone Keyer input to the same level as when only one tone is present. The Side-tone Keyer is connected to the 1.5 kHz oscillator only, as TR12 is turned off by means of D24.

When the TUNE-button of the Transmitter Power Amplifier is activated, terminal 18c is LOW. A2H is selected by means of $D 1$ and D3, the transmitter is keyed via IC2 pin 9, and the Side-tone Keyer is turned off by means of $I C, 9$ pin 1.

### 5.4.3 /235 HARMONIC GENERATOR

There is one phase looked loop and its associated lock detector located on this printed circuit board.

The reference frequency of 1.4 MHz for this loop is derived from the TCXO, in order to accomplish the necessary degree of frequency stability of the output signal from the Voltage Controlled Oscillator.

The fixed reference frequency of 1.4 M Hz is fed to one of the two input ports of the Phase / Frequency Detector. The output from the loop Divider is fed, via IC2, to the other input port of the Phase / Frequency Detector, which in the case of a frequency difference between the two signals will produce a DC-error voltage.

The purpose of the Low Pass Filter is to remove the pulses from the output of the Phase/Frequency Detector, thus allowing only the DC-error voltage to pass on to the vari-cap diodes of the Voltage Controlled Oscillator.
The outputs of the VCO are amplitude regulated and fed to the RFTRANSLATOR /233\ and Pre-scaler respectively.

The division ratio of the Pre-scaler and the Loop Divider is controlled by four outputs of the Band Decoder, IC4, which is again controlled by the BAND-information stored in the MEMORY /238\.

Four other outputs of the Band Decoder select, via the Band Selector IC1, the proper band of the $V C O$ and the proper band pass filter of the RF TRANSLATOR /233\.

The Lock Detector detects any frequency difference between the two signals fed to the input ports of the Phase/Frequency Detector. When a frequency difference is observed, the Lock Detector turns on TR9, thus pulling down the KEYLINE to a low level voltage, as long as frequency equality is not accomplished.

The Loop Filter and Voltage Controlled Oscillator are located on this printed circuit board.

The Loop Filter consists of an active4th order low pass filter and an integrator. The purpose of the Loop Filter is to remove the pulses from the output of the Phase/Frequency Detector on board /237\ and allow only the DC-information to pass on to the vari-cap diodes of the VCO. The VCO is amplitude stabilised.

The selection of one of the three bands of the VGO is carried out by means of a decoding circuit on /237\.

### 5.4.6 /237\ FREQUENCY DIVIDERS A

There are two divider chains, their associated buffer amplifiers, a 2182 KHz pre-selection circuit, a lock detector, and a phase 1 frequency detector located on this printed circuit board.
The Reference Divider divides the 1.4 MHz from the TCXO by 1400 , thus providing the 1 KHz reference frequency for the Phase/Frequency Detector.

The Loop Divider chain consists of four programmable, decimal counters, IC16, IC20, IC23, and IC26, which are loaded at the end of each counting cycle with the data blocks supplied by the Memory /238\ or the 2182 kHz Pre-selection circuit. The Side-step Control consists of a decimal counter, IC6, a four bit comparator, IC10, and gating circuitry. A data block, containing the information about the 100 Hz setting is fed to the comparator. At the end of each counting cycle IC6 is incremented by one and its outputs (B) are compared with the content of the data block (A). As long as $B$ is less than $A$ the division ratio of the four programmable counters are equal to the BCD 9's complement of their four associated data blocks. When $B$ becomes equal to or greater than $A$ the division ratio is incremented by one. This change of the division ratio is controlled by output pin 5 of IC10, which selects one of the two decoding branches, branch 1 and branch 2, consisting of parts of IC3, IC4, ICS, and IC8, and one half of IC1.

When the mode Fl, Al, or A 2 H is chosen, the division ratio is decreased by one and two alternately, independent of the above mentioned side stepping procedure. This is controlled by the "1.5 kHz SHIFT" line and performed by the other half of IC1, where output pin 9 is changed at the conclusion of each counting cycle, thus shifting between the two decoding branches, branch 3 and branch 4, alternately. This means that the division ratio is decreased by 1.5 on the average. Because the reference frequency of the loop is 1 kHz , this means that the output frequency of the loop is decreased by 1.5 kHz .

The output from the Loop Divider chains fed to the Phase / Frequency Detector and is here compared to the fixed 1 KHz reference frequency. In case of a frequency difference the detector will produce a DC-error voltage which will correct the VCO en board /236 to establish the wanted frequency equality.

The two inputs /238 $-9 a$ and /238 MHz setting is decoded in order to select the proper band of the VCO on board /236. If both of the two inputs are HIGH, the Disable output line is activated.

The Look Detector consists of IC15 and IC17. If the loop is in look, output pin 9 of IC19 will always be LOW when the output pulse from the Loop Divider is generator, so output pin 9 of IC15 stays LOW.

In an unlocked condition of the loop, output pin 9 of IC19 will be at a LOW or HIGH logic level at random when the output pulse from the loop Divider is generated. The first mono-stable multivibrator of IC17 is triggered by the corresponding output changes of IC15, thus, via the last half of IC17, inhibiting the Keyline.

If the reference signal of the loop is missing, this is detected by the last half of IC17, again inhibiting the Keyline.

### 5.4.7 /338\ MEMORY

This printed circuit board contains 24 sockets, organised as eight columns (No. one to No. eight) by three rows (A, B, and C). Thus up to 24 Programmable Read Only Memories (PROMS) corresponding to 256 different Frequency Nos. can be mounted in the MEMORY.

The PROM used here is organised as 32 words by 8 bits. Because 24 bits are needed as control information for each Frequency No. three PROM's must be driven in parallel, corresponding to an extended memory block organised as 32 words by 24 bits.

A memory block consisting of three $P R O M^{\prime}$ s is mounted in one of the eight columns. The $P R O M ' s$ in column No. one contain the necessary 24 bits associated with the Frequency Nos. from 0 to 31, column No. two contains the 24 bits associated with the Frequency No's from 32-63 and so on (see table 3.12.1).

From board /239\ the binary encoded Frequency No. selects, via 5/6 IC1, one of 32 words and, via IC2, one of eight columns.

### 5.4.8/339\ DISPLAY AND KEYBOARD

When a key is depressed the associated key-number is encoded to the BCD code by means of IC2 and IC4. This BCD coded data block is fed to the input ports of the first register IC8 of the Digit Register Stack.

When a key is depressed or released some sort of bouncing effect will always appear before the key has settled. This bouncing is removed by means of the Key Bounce Eliminator, consisting of IC5 and the associated extern at components. A clock pulse is produced when the key has settled, after being activated, and then the data block mentioned above is read into the first register of the Digit Register Stack, and the data blocks already stored in this Stack are shifted to the next register. The outputs of the Digit Register Stack are the BCD code of the Frequency No. chosen. Via the three BCD to 7-segment decoderdrivers, IC12, IC14, and IC19, the Frequency No. is displayed. The BCD encoded Frequency No. is converted to its associated binary code before it leaves this board, by means of IC15, IC16, IC20, and IC21, in order to select the proper information stored in the Memory /238\. ]t is not possible to choose a Frequency No. greater than 255 , if such an attempt is done a decoding circuitry of $3 / 4$ IC7 automatically clears the Digit Register Stack.

By means of the variable voltage regulator consisting of TRI, TR2, and associated components the three Led-displays can be dimmed continuously.

When the BAND switch is turned into the 2182 kHz position, terminal 11 receives a LOW level signal which, via 3/4 IC6, activates the ripple blanking outputs of IC12, and IC19. The displays are then blanked, but the original Frequency No. is stored.

If a HIGH level is applied to terminal 13 (Disable) or a LOW level to terminal 14 (Programmed Frequency No.), the Blanking Oscillator consisting of IC3 and associated components starts a 1 Hz oscillation and the Key-line is inhibited. Thus the displays, via 3/4 IC6 and IC12, IC14, and IC19, will start to flash with a frequency of 1 Hz . However, if the BAND switch is turned to its 2182 kHz position, the levels of terminals 13 and 14 do not affect the Key-line.
5.4.9 /240\ MASTER OSCILLATOR

The oscillator itself is a sealed unit containing a highly stable temperature compensated crystal oscillator, TCXO, at 11.2 MHz . The output signal of the TCXO is amplified in the transistors TRI to TR3 and fed to the binary counter IC1 that divides the input frequency by 8. The 1.4 MHz square wave signal is filtered in a tuned Circuit C6, T1, and the re-suiting sine wave signal is fed to the output terminals.

TCXO's of two different manufactures may be used. In both cases crystal oscillator ageing is very small (less than $10-11$ per anum) and will be greatest during the first few years. Ageing will normally cause an increase in frequency, which in the one case can be compensated for by introducing the connection indicated by the dotted line in the circuit diagram (this will reduce the frequency by approx. 2*10-6) and in the other case by changing the factory selected resistor.

The resistor should be selected at $25^{\circ} \mathrm{C}$ ambient temperature to give a TCXO output frequency offset from the nominal frequency (11.2 MHz) by the amount marked on the can.

Frequency adjustment should be carried out only if a high quality counter is available for control of the frequency. It must be ensured that the accuracy of the counter at the time of use is better than 10-7.

### 5.4.10 /241

The purpose of this printed circuit board is to make it possible to program manually the $P_{R O M}$ 's used as building blocks in the MEMORY /238\.

The only types of $P R O M^{\prime} S$, which can be programmed by means of the PROGRAMMING UNIT are the TI-types shown in table 3.11.1. Three PROM's together can contain the information of 32 different Frequency $\mathrm{No}^{\prime} \mathrm{s}$. The three $P R O M$ 's are placed in SK1, SK2 and SK3. Their input word address is chosen by means of the key-board on the front panel, thus selecting one of the 32 possible words in each of the three PROM'S.

A programmed output will be greater than $2 V$ (HIGH), an un-programmed output will be less than 0.8 V (LOW). To prevent an attempt being made, by mistake, to program new information into a word which has already been programmed once, one bit location, pin 9 of the PROM placed in SK1, must always be programmed if the rest of the bit locations of a Frequency No. have been correctly programmed.

Half of IC4 is always sensing the voltage level on pin 9 of SK1. If this voltage level is HIGH, pin 5 of IC4 will also become HIGH after the first positive transition of the clock pulse at pin 3, thus disabling IC2 and thereby IC6 and IC8 from being activated by the key S1.

By activating the key S2, the output of pin 9 of SK1 will be forced LOW, as for an un-programmed Frequency No., thus permitting the user via the key S1 to activate IC and there by IC6 and IC8.

Their outputs are combined by means of $4 / 6$ IC5, $2 / 4$ IC7 and 1/4 IC3 into three pulse-trains, one for the Vcc-pins and one for the Enable pins of the three $P R O M$ 's and lastly one for the transistor TR1. This transistor sinks the programming current from the PROM-output to which the Programming Pin is connected.

The three pulse-trains are shown below.

The two voltage levels of the Vcc-pins of the PROM's are stabilised by means of D3 and IC9.
 to obtain good programming re-suits, their Vcc supply line only receives a 5 V pulse in 5 msec out of 100 msec during the period of time where no programming takes place. This pulsed operation is controlled by a clockpulse generated by IC1 openin, 0 and closing, via $2141 G 3$ and 216 ICS, the transistorTR2.

At the end of each 5msec period the logic levels of pin 9 of SK1 and the Programming Pin are read into the two D-flipflops of IC4; a HIGH level, corresponding to a programmed bit location will make the associated lamp light.

When the key SK1 is activated and MON1 of IC2 is triggered, the clock generator IC1 is stopped. Once the programming pulse-trains have been accomplished the clock generator is allowed to start again after a delay of approximately 30 msec .

A delay circuit consisting of R71, R72, C12 and D4 prevents the Vccprogramming pulse from being generated when the PROGRAMMING UNIT is first switched on. This prevents false generated signals to IC2. This is accomplished by turning off TR2 via 1/4 IC3 and 2/6 IC5 until all the voltage levels have stabilised.


### 5.5 Circuit Summary, Transmitter Power Amplifier T 5000

5.5.1 The circuit diagram is divided into a wiring diagram On page 8-89 showing the interconnections between the printed circuit boards of which the Transmitter Power Amplifier is composed, and circuit diagrams of the individual circuit boards. The block diagram on page 8-61 illustrates the operation of the Transmitter Power Amplifier.
5.5.2 The RF signal from the Exciter is amplified in the DRIVER /251 and applied to the P.A. output valves. Antenna tuning is pre-selected with the BAND switch, and fine tuning is performed with the TUNING control.
5.5.3 Antenna current measuring and CHECK SWITCH circuits are contained on printed circuit board /252\.
5.5.4 Visual indication of whereto set the BAND-switch in accordance with the frequency selected is performed by the BAND INDICATOR /253\. When 2182 kHz is selected, information is supplied to the Exciter. If the BAND switch is incorrectly set, the Key-line inhibit output prevents keying of the transmitter.
5.5.5 Regulation of grid and screen grid voltages, as well as filament voltages $f$ or the P.A. output valves, is carried out on printed circuit board /254\.
5.5 Circuit Description, Transmitter Power Amplifier T5000 DRIVER
5.6.1 /251 DRIVER

A preamplifier consisting of TR1 and TR2 in parallel amplifies the signal which is then applied to a class-a push-pull stage consisting of the transistors TR3 and TR4. The DC current of these transistors is set by means of the variable resistor $R 7$. Coarse setting is by means of the resistors R13-R15. The compensation network at the output provides a flat response over the frequency range.
5.6.2 /252\ METER CIRCUIT

The printed circuit board is an integral part of the CHECK SWITCH and contains voltage dividers for metering different voltages of the transmitter Power Amplifier. Dl rectifies the current of the antenna current transformer. D2, R11, and R13 provide compression of the meter scale. TR1 inverts the negative grid voltage to a positive current in the meter when checking VG.
5.6.3 /253\ BAND INDICATOR

The purpose of the Band Indicator is to show where to set the BAND switch in accordance with the frequency selected. This is done by means of 16 light emitting diodes, 15 of which are controlled from two 1 of 10 decoders IC3 and IC5. These decode the binary band information from the Exciter and make one of the lamps flash if the BAND switch is incorrectly set. The BAND switch information and the information from the Exciter are compared in the Comparator, IC4. If they are not equal the output will be LOW and will via lC6and IC7, start the 1 Hz Muitivibrator (FLASH ENABLE: HIGH) that controls the decoders via 2/4 IC1. At the same time the Key-line output is Low thus inhibiting transmission.

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When "2182 kHz" is selected by means of the BAND switch the information is HIGH on al] four input leads from the BAND switch. In that case the output from 1/2 IC2, pin 8, is HIGH and as terminal 7 is open and thus HIGH too, the output of $1 / 4$ IC1, pin 11, is LOW, telling the Exciter to select 2182 kHz . The light emitting diode Dl will show constant light.

### 5.6.4 /254 V VOLTAGE REGULATORS

The Grid Voltage Regulator consists of two integrated circuits IC1 and IC2 that operate as zener diodes having a very low temperature coefficient. The actual grid voltage is set by means of the variable resistor R1.

The screen grid voltage of the power amplifier tube Vl is fixed and determined by two 150 V zener diodes 250 Dl and 250 D3. In order to be able to equalise the currents of the two tubes, the screen grid voltage of $V 2$ is made adjustable. The zener diodes 250 D5 and 250 D6 are connected to a voltage-regulator containing transistor 250 TR1, the collector voltage of which can be set by means of variable resistor R3.

The 6V filament voltages of, the power amplifier tubes has to be controlled within very narrow limits to obtain maximum tube life. This is done in the Filament Voltage Regulator. In order to reduce current consumption the filaments of the two tubes are series connected, which means that the voltage of each tube has to be control- led. The regulator of $V 2$ consists of the NPN darlington transistor 250 TR1 in connection with the zener diode $D 4$ and resistor R15. When the voltage of the filament reaches 6 V , the transistor starts to conduct and shunts the filament by which the voltage is kept constant. The regulator of $V$, is identical except the transistor is a PNP darlington. The collector currents of the two transistors pass the diodes D5 and D6, and the voltage across these diodes is applied to a current regulator consisting of TR1 and 250 TR2. TR3 serves to limit the peak current.

### 5.7 Circuit Summary, 24V DC Power Pack P5000

5.7.1 The circuit diagram is divided into a wiring diagram on page 8-99 showing the interconnections between the printed circuit boards of which the Power Pack is composed, and circuit diagrams of the individual circuit boards. The block diagram on page 8-62 illustrates the operation of the Power Pack.
5.7.2 Filament supply for the transmitter power amplifier output valves is obtained directly from the battery, while the remaining necessary supply voltages for transmitter Power Amplifier, Exciter and Receiver are generated in converters. Converter transformers give full isolation between the battery voltage and the chassis, permit- ting these to be earthed without causing the supply leads to be earthed.
5.7.3 Low tension voltages for Exciter and Receiver and negative grid voltage for the Transmitter Power Amplifier are generated by the LT-converter followed by rectifiers and stabilisers.

The LT-converter frequency is determined by an RC-coupled oscillator followed by a bi-stable muitivibrator.
5.7.4 High tensions for the P.A. valves are generated by the HT-converter followed by rectifiers. To avoid excessive anode voltage to the P.A.valves at high battery voltages an input voltage sensor is incorporated in the power pack. When the input voltage exceeds a certain voltage a relay in the input voltage sensor circuit will operate and decrease the anode voltage.

Supply voltage $f$ or the driver stage in the transmitter power amplifier is also generated in the HT-converter followed by rectifiers and a voltage stabiliser.

The HT-converter is driven from two secondary windings on the LTconverter transformer.
5.7.5 To fulfil the input voltage requirements of the blower in the transmitter power amplifier, a separate blower converter is incorporated.

### 5.8 Circuit Description, 24V DC Power Pack P5000

5.8.1 /260 LT -, HT-, AND BLOWER-CONVERTER

Relay RL1 serves as a mains switch. The current which energises RL1 is controlled by the POWER switch. Turning the switch from OFF to RECEIVE ONLY connects the relay coil to the battery voltage. Diode 261 Dl ensures that the relay cannot operate if the battery voltage does not have the correct polarity. The contacts of RL1 are protected against excessive capacitor charging current by resistor R2. The resistor ensures that the capacitors after the relay are always charged before the contacts are closed.

The LT-converter is composed of transistors TR10 and TR12, and the square-wave output signal is coupled through transformer T3 to the LTand grid voltage-rectifiers and as a drive signal to the HT-converter.

The HT-converter contains transistors TR4, TR6, TR8 and TR11 and the output signal is coupled through transformer $T 2$ to the HT-rectifiers.

The Blower-converter contains transistors TRI and TR2 and the output transformer $T 1$ transforms the square wave output voltage to the correct level.

Included in the Power Pack is also a loudspeaker connected to the associated receiver AF-output.
5.8.2 /261 CONVERTER DRIVER LT-RECTIFIERS AND LT-STABILIZERS.

A bi-stable muiti-vibrator composed of transistors TR2 and TR4 is driven by oscillator IC1 with a nominal frequency of 440 Hz . The output signals from the muiti-vibrator are therefore square waves with a repetition frequency of 220 Hz , and these signals having a phase difference of 180 " are, after current amplification in transistors TR1 and TR3, used as driving signals for the LT-converter.

One output signal from the LT-converter is rectified in diodes D11 and D12 which are followed by two voltage stabilisers both having a nominal voltage of 7.5 V .

Another output signal from the LT-converter is rectified in diodes D16 and D17 followed by a voltage stabiliser having a nominal output voltage of 15.7 V .

The board contains also the keying relay RL1 for the complete TRP5000 equipment. The contacts of the relay are used to disconnect the speaker built into the Power Pack, and to key the grid bias of the P.A. valves.

## 5.8 .3 <br> /262\ HT-RECTIFIERS AND BASE KEYING CIRCUIT.

The current which energises relays RL1 and RL2 passes Transmitter Power Amplifier safety switch 250 S1 and Power Pack safety switch 260 S2 and is controlled by keying relay 261RL1. Via the contacts of RL1 and RL2, base drive is applied to the HT-converter transistors and the square wave output is rectified in six bridge circuits. The DC-outputs of each bridge are series connected to obtain the anode supply voltage for the Transmitter Power Amplifier, while the DC-output from the second bridge above ground is used as screen grid supply voltage.

When the MODE switch of the associated Exciter is set to Fl, the anode voltage is lowered by opening the contacts of relay RL3 and thereby the anode dissipation in the Transmitter Power Amplifier is de-creased.

The anode voltage is further controlled by relay 264 RL 1 which in the case of high battery voltage will de-crease the available anode-voltage by disconnecting the input voltage to one of the six bridge rectifier circuits.

### 5.8.4 /263\ DRIVER- AND GRID-VOLTAGE STABILIZERS, BLOWER CONVERTER DRIVER AND TIMER

An output voltage from the HT-converter transformer 260 T 2 is fed to rectifiers Dl and D2 followed by a voltage stabiliser having a nominal output voltage of 28 V .

An output voltage from the IT-converter transformer 260 T 3 is fed to bridge rectifier D3, D4, DS and D6 followed by a zener diode stabiliser. The stabilised voltage is negative with respect to chassis and is further stabilised in the Transmitter Power Amplifier to be used as grid voltage.

To avoid the transmitter being keyed before the filaments of the P.A. valves have reached their operating-temperature, a 30 seconds timer is included. The timer is composed of integrated circuit 101 and will commence the timing period when the POWER switch 260S1 is turned from RECEIVE only to STAND BY. Until the 30 seconds period has elapsed the keying relay 26 RL1 con not be operated.

The blower in the Transmitter Power Amplifier requires a 50 Hz input voltage. A bi-stable muitivibrator composed of transistors TR2 and TR4 is driven by oscillator TR1 with the nominal frequency 100 Hz . The output signals from the muitivibrator are therefore square waves with a repetition frequency of 50 Hz , and these signals which have a phase difference of $180^{\circ}$ are used as driving signals for the Blower-converter.

### 5.8.5 /264\ INPUT VOTAGE SENSOR

A Schmitt-trigger having a hysterisis of approx. 2 volts is used as an input voltage sensor. If the battery input voltage exceeds a certain value relay RL1 will open and decrease the anode voltage by opening one of the secondary windings on the HT-transformer 260 T 2 . When the battery voltage is approx. 2 volts below the trigger point relay RL1 will close, increase the anode voltage and thereby ensure that the anode dissipation of the P.A. valves are within safe limits.

The -board also carries a noise suppression filter for the Blowerconverter.

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### 5.9 Circuit Summary, AC Power Pack P5001

The circuit diagram is divided into a wiring diagram on page 8-109 showing the interconnections between the printed circuit boards of which the Power Pack is composed, and circuit diagrams of the individual circuit boards. The block diagram on page 8-63 illustrates the operation of the Power Pack.

All necessary supply voltages for Transmitter Power Amplifier, Exciter and Receiver are obtained through transformers $T 1$ and $T 2$. The transformers give full isolation between the mains input voltage and chassis, permitting these to be earthed without causing the mains leads to be earthed.
Low tension voltages for Exciter and Receiver and negative grid voltage for the Transmitter Power Amplifier are obtained through transformer T1 followed by rectifiers and stabilisers.

Transformer $T 1$ also carries secondary windings for filament supply to the P.A. valves and for the blower in the Transmitter Power Amplifier.

High tension for the Transmitter Power Amplifier is obtained through transformer T2 followed by rectifiers.

Supply voltage for the driver stage in the Transmitter Power Amplifier is also obtained from transformerT2followed by rectifiers and a voltage stabiliser. Included in the Power Pack is also a loudspeaker connected to the associated receiver AF-output.

### 5.11 Circuit Description, AC Power Pack P5001

5.10.1 /265 TRANSFORMERS

A double mechanically operated safety switch is inserted in the mains input leads. The primary current of transformer TI is controlled by the POWER switch. Turning the switch from OFF to RECEIVE ONLY connects the transformer to the mains and low tension is supplied to the Receiver. Relay RL1 connects the primary windings of transformer $T 2$ to the mains.
5.10.2 /266 LT-STABILISERS

One output voltage from transformer 265 T 1 is rectified in the chassis mounted diodes 265 D1 and 265 D2. Two voltage stabilisers, one supplying the Receiver and one the Exciter, have a nominal voltage of 7.5 V .

Another output voltage from transformer 265 T is rectified in diodes D8 and D9 followed by a voltage stabiliser having a nominal output voltage of 15.7 V . This voltage is supplied to both Receiver and Exciter.

The board also contains the keying relay RL2 f or the complete TRP5000 equipment. The contacts of the relay are used to disconnect the speaker built into the Power Pack and to key the grid bias of the P.A. valves.

### 5.10 .3 <br> /267\ HT-RECTIFIERS A

HT-transformer 267 T2 has six secondary HT-windings. Each winding is connected to a bridge rectifier and the DC-outputs are series connected to obtain the anode supply voltage for the Transmitter Power Amplifier, while the DC-output from the second bridge above ground is used as screen grid supply voltage.

When the MODE switch of the associated Exciter is set to Fl, the anode voltage is lowered by Opening the contacts of relay RL1 and thereby the anode dissipation in the Transmitter Power Amplifier is decreased.
5.10.4 /268 DRIVER AND GRID VOLTAGE STABILIZERS

An output voltage from the HT-transformer 265 T 2 is fed to rectifiers Dl and D2 followed by a voltage stabiliser having a nominal output voltage of 28 V .

An output voltage from the transformer 265 Tl is fed to bridge rectifier D3, D4, D7 and D8 followed by a zener diode stabiliser. The stabilised voltage is negative with respect to chassis and is further stabilised in the Transmitter Power Amplifier to be used as grid voltage.

To avoid the transmitter being keyed before the filaments of the P.A. valves have reached their operating temperatures a 30 seconds timer is included. The timer is composed of integrated circuit $1 G 1$ and will commence the timing period when the POWER switch 265 S2 is turned from RECEIVE ONLY to STANDBY. Until the 30 seconds period has elapsed the keying relay cannot be operated.

The board contains also a filament-voltage rectifier D5 and D6, and relay RL1 is used to apply power to the blower in the Transmitter Power Amplifier.

## 6. Simple Service

### 6.1 Incorrect Operation

If the equipment is not functioning correctly, a check should be made that it is being operated properly. Go through adjustment procedures 2.2 and 2.3 if necessary.

### 6.2 Battery

The condition of the battery should be checked at frequent intervals. The battery must always be fully charged and should be topped up frequently with distilled water (liquid should rise 5 to 10 mm above the plates).

### 6.3 Checking the Antenna Tuning

The antenna tuning may be checked by adjusting the transmitter as described in section 2.3, preferably on one of the frequencies that was listed in the TUNING CHART when the equipment was installed. Then set POWER switch to SIMPLEX, MEDIUM and CHECK SWITCH to LEVEL. Pull the CHECK SWITCH knob out and press TUNE button. Note the meter reading. Set POWER switch to SIMPLEX, FUIL and press TUNE button. Note ANTENNA CURRENT reading.
Compare the readings with the readings listed in the TUNING CHART. ]f the two readings differ appreciably from the listed values, and the transmitter is otherwise functioning normally, the fault can be expected to be in the antenna system or in the transmitter earth connection.

Accordingly, the following cheeks should be made:
Have any changes been made in antenna or earth connections since the installation was made?

Have any changes been made in the rigging of the vessel, or in the placement of derricks etc.?

Is leakage present on the antenna, possibly caused by moisture or dirt on the antenna insulators?

### 6.4 Using the CHECK SWITCH

The CHECK SWITCH is not normally operative. Pulling the switch knob out will switch the ANTENNA CURRENT meter to read the voltage or current selected with the switch. When released, the knob will return to its original position.

The meter reading in all positions except LEVEL is approx. 3 under normal conditions, i.e. transmitter adjusted as described in section 2.3, POWER switch set to FULL and TUNE button pressed.

| Position of CHECK SWITCH | Check of | Deflection to 3 corresponds to approx. | Actual deflection |
| :---: | :---: | :---: | :---: |
| VA | anode DC voltage | 1700 V | ```2.4 to 3.5 dependent on supply voltage``` |
| Vs1 | screen grid voltage of valve no. 1 | 300 V |  |
| VS2 | screen grid voltage of valve no. 2 | 300 V | 2.4 to 3.6 dependent on valve characteristics |
| VG | control grid bias | -49V | 2.3 to 3.7 dependent on valve characteristics |
| I1 | cathode current of valve no. 1 | 165 mA |  |
| I2 | cathode current of valve no. 2 | 165 mA |  |
| VD | supply voltage to driver amplifier | 28 V |  |

Table 6.4.1

If the meter shows no reading in these positions, reference should be made to section 6.5.
With POWER switch set to DUPLEX, FULL and TUNE button not pressed, valve current readings, I1 and I2, in the A3J mode, un-modulated will be approx. 1.4, and in the A3H mode, un-modulated approx.2.3.
6.5 Replacement of Fuses

All fuses, except the high tension fuse, are accessible on the front panel of the Power Pack. The high tension fuse becomes accessible when the Power Pack is pulled out.

Note: Set SUPPLY switch to OFF and open external supply voltage switch before opening the equipment and replacing fuses. Short circuit both ends of high tension fuse to chassis using an insulated tool before touching it.

Fuse ratings are given in table 6.5.1 and table 6.5.2 below. Fuses with marked ratings within 5 per cent of the ratings given must be used. Note that slow or fast blowing fuses must be used where specified.

| Fuse Rating | Designation | Front Panel Fuses (from left to right) | Symptom if fuse is blown |
| :---: | :---: | :---: | :---: |
| 1.6A fast | 260FS 5 | $7-5 v$ to Receiver | no light in display |
| 1.6A fast | 260FS 6 | 15-7V to Receiver and Exciter | ```no light in displays, negative deflection in CHECK SWITCH position Vg``` |
| 3.15A fast | 260FS 3 | 7-5V to Exciter | no light in displays |
| 40A fast | 260FS 2 | 24 V battery input <br> (LT-and HT-converters) | no light in displays |
| 8A fast | 260FS 1 | ```24V battery input (RL1, filament supply, blower)``` | no light in displays, no cabinet light |
| 1.6A fast | 260FS 4 | 28 V to Driver in T5000 | no reading in CHECK SWITCH position VD |
| 0.5 A fast | 260FS 8 | 600V screen grid supply | no reading in CHECK SWITCH positions Vs1 and Vs2 |
| 0.5 A HT | 260FS 7 | High Tension Fuse 1700 V to anodes | no reading in CHECK SWITCH position Va |

Fuse Ratings, 24V-DC Power Pack P5000
Table 6.5.1.

| Fuse Rating | Designation | Front Panel Fuses <br> (from left to right) | Symptom if fuse is blown |
| :---: | :---: | :---: | :---: |
| 1.6A fast | 265FS 3 | $7-5 v$ to Receiver | no light in display |
| 1.6A fast | 265FS 7 | 15-7V to Receiver and Exciter | no light in displays, negative deflection in CHECK SWITCH position Vg |
| 3.15A fast | 265FS 2 | $7-5 \mathrm{~V}$ to Exciter | no light in displays |
| 8/16A slow | 265FS 1 | mains input | no light in displays no cabinet light |
| 8A fast | 265FS 8 | filament supply, | no reading in CHECK SWITCH positions II and I2 |
| 1.6A fast | 265FS 6 | 28 V to Driver in T5000 | no reading in CHECK SWITCH position VD |
| 0.5 A fast | 265 FS 5 | 600V screen grid supply | no reading in CHECK SWITCH positions Vs1 and Vs2 |
| 0.5A HT | 265FS 4 | High Tension Fuse 1700 V to anodes | no reading in CHECK SWITCH position Va |

Fuse Ratings, AC Power Pack P5001

### 7.1 Introduction

Repair and adjustments on the equipment shout be preformed only by qualified technicians, to whom this chapter is addressed. Before attempting any repairs or adjustments, a study of Chapter 5, Technical Description, is recommended.
7.3 Locating Subunits and Components

Locations of circuit boards in the equipment appear from the photographs on pages 8-56 and 8-59. Locations components on each circuit board appears on the component location drawings against the respective circuit diagrams.

### 7.4 Locating Faults

Fault findings, as described in section 7.5 below, is aided by test points provided for the purpose of permitting rapid localization of faulty circuit boardson the basis of DC measurements. Since not all types of faults can be traced by means of DC measurements, supplementary AC measurements with an oscilloscope may be required; see section 7.6 . To facilitate fault finding on each individual circuit board, typical voltages are listed on the circuit diagrams.

## 7.5

## Test Points

Several circuit boards contain one or more test points. They are small pin-type terminals, colour coded following the standard colour code in addition to being numbered. In the circuit diagrams, test points are maked [TP1], [TP2] etc., and typical voltages at the test points are listed there.

The terminals of the circuit boards may to a great extent also regarded as test poits. Tipical voltages are therefore also listed against relevant terminals on the circuit diagrams.

If a voltage measured at a test point differs markedly from the listed value it is faily certain indication that the circuit board in question is faulty, assuming that the voltages applied to the circuit board are the correct ones. This should likewise be checked.

7-1

AC voltages listed in the circuit diagrams are typical voltages. Voltages specific are based on measurement with an oscilloscope having an input impedance of $10 M o h m s$ in parallel with $7 p F$, a sensitivity of the order of $50 \mathrm{mV} / \mathrm{div}$ and a frequency range of not less than $\mathrm{DC}-50 \mathrm{MHz}$.

AC voltage values measured in the signal path of the transmitter can be measured only if the transmitter is modulated with a two-tone signal. This can be done by pressing the TUNE button of the transmitter.

### 7.7 DC Voltages

DC voltages listed in the circuit diagrams are based on measurement with a $25 \mathrm{kohms} / \mathrm{volt}$ multimeter. If a stated voltage is dependent on the setting of a control, this is also stated on the circuit diagrams. Typical logic levels (LOW/HIGH) are indicated in brackets.

### 7.8 Replacement of Valves

7.8.1 Replacement of one or both of the valves in the transmitter power Amplifier normally requires re-adjustment of the quiescent current and RF drive voltage to the valves.

Note: Set SUPPLY switch to OFF before opening the equipment. Short anodes of valves to chassis, using an insulated tool, before touching the valves.

The adjustment procedure following a valve replacement must be performed as follows.
7.8.2 Adjustment of Quiescent Current:

The quiescent current of each valve must be 75 mA and may be measured by means of the built-in meter. Readings in position I1 and I2 of the CHECK SWITCH (pulled) should be 1.4 in the A3J-mode, un-modulated. The adjustment is made by means of the trimming potentiometers 254R1 and 254R3 located behind the two membane grommets in the upper left corner of the valve chassis. The left hand potentiometer controls the screen grid voltage of valve No.2. The right hand potentiometer controls the grid bias of both valves. Use a 2 mm wide screw driver for the adjustments.

1. Pull the Transmitter Power Amplifier unit out as far as the built-in stops permit, and pull the piston of the safety switch outwards to the locked position.
2. Set SUPPLY switch to TRANSMIT, POWER switch to SIMPLEX and MODE switch to A3J.
3. Key a Frequency No. programmed to a radiotelephone frequency into the keyboard and set BAND switch accordingly.

30 seconds after switching to TRANSMIT the transmitter can be keyed. Keying is performed by setting the POWER switch to DUPLEX or by pressing the handset key. If the handset key is used, remove the microphone capsule in order to avoid unwanted modulation of the transmitter.

Note: Extreme care must be taken as high tension is now accessible.
4. Key the transmitter and adjust $\mathbf{I}_{1}$ by means of 254 R1 (right) to 1.4 on the meter scale.
5. Key the transmitter and adjust $\mathbf{I}_{2}$ by means of 254 R3 (left) to the same value.
6. Keep the transmitter keyed for a period of 5-10 minutes and repeat points 4.and 5. if necessary.
7.8.3 Adjustment drive voltage:

Due to the spread in valve characteristics it may also be necessary to re-adjust the grid drive to the valves.
Adjustments are made separately for the CT-band (1.6 to 4 MHz ) and each HF-band by means of the Level Setting potentiometers on printed circuit board /233\} in the Exciter. These become accessible by removing the lid of the screened can below the chassis.

The potentiometers are adjusted so that a maximum valve current of between 165 mA and 175 mA is obtained per valve with POWER switch set to FULL and TUNE button pressed. This corresponds to a deflection of between 3.0 and 3.2 on the built-in meter with the CHECK SWITCH (pulled) in position 11 and 12. Note that maximum current is obtained when the ANTENNA TUNING is de-tuned. If the two currents vary slightly, the highest reading must not exceed 3.2.

## 7.9 <br> Adjustments

The following sections describe alignment procedures for printed circuit boards that contain adjustable components. Bear in mind that no adjustments should be carried out unless there is a clear indication that it is really necessary. Moreover, adjustments should be carried out only by qualified technician with the necessary equipment at his disposal. When a unit or printed circuit board is replaced, adjustments are in some cases necessary. These cases are listed in the table blow.

| Replacement of <br> unit or board | Adjustment <br> required of | Procedure given <br> in section |
| :--- | :--- | :--- |
| E 5000/E 5001 | Level Setting | 7.8 .3. |
| $\underline{1232 \backslash}$ | $232 R 15$ | 7.10 .2 |
| $\underline{1233 \backslash}$ | Level Setting | 7.8 .3 |
| $\underline{T 5000}$ | $232 R 15$ | 7.10 .2. |
|  | Level Setting | 7.8 .3 |
|  | Antenna Tuning | 3.14 |

7.10 /232\ Realignment of 1.4 MHz (1.973 MHz) EXCITER

Measuring equipment:

Oscilloscope having a sensitivity better than $50 \mathrm{mV} / \mathrm{div}$. Input impedance 10 Mohm in parallel with 20 pF or less.
7.10.1 Realignment of 232R6, Microphone Sensitivity:

The microphone sensitivity potentiometer is normally fully clockwise, corresponding to full sensitivity. When the transmitter is installed where there is a high acoustic noise level it can be advantageous to reduce the sensitivity by turning the potentiometer half or fully anticlockwise. This has the effect of reducing the back- ground noise coming up in speech pauses. The compressor ensures that the transmitter is still fully modulated by the speech signal.
7.10.2 Realignment of 232R15:

Control settings: SUPPLY switch: STAND BY

1. Connect oscilloscope to test point $232 T P 7$.
2. Unsolder brown lead to terminal 232-4.
3. Set potentiometer 232 R 15 to the middle of its range.
4. Press TUNE button and adjust sensitivity of oscilloscope to give a full screen peak to peak deflection.
5. Re-solder brown lead to terminal 232-4.
6. Press TUNE button and adjust 232 R15 to give exactly the same deflection on oscilloscope as before.
7.10.3 Realignment of 232R42, Balance:

Control settings: SUPPLY switch: STAND BY
MODE switch: A3J
BAND switch-: not 2182 kHz

1. Connect oscilloscope to test point 232 TP6
2. Adjust potentiometer 232 R 42 for minimum deflection on oscilloscope.
7.10.4 Realignment of 232R46, Carrier Level

Control settings: SUPPLY switch: STAND BY
MODE switch: A3H

1. Connect oscilloscope to 232c42 / 232L6
2. Press TUNE button and adjust sensitivity of oscilloscope to give a full screen peak to peak deflection.
3. Un-solder grey lead to terminal 232-13.
4. Press TUNE button and adjust 232 R 46 to give exactly the same peak deflection as before.
5. Re-solder grey lead to terminal 232-13.

Above procedure implies that 232 R15 is correctly adjusted (point 7.10.2.).
7.10.5 Realignment of 232T1:

Control settings: SUPPLY switch: STAND BY

1. Connect oscilloscope to test point $232 T P 7$.
2. Press TUNE button and adjust 232 T1 for maximum deflection on oscilloscope.
7.10.6 Realignment of 232L6:

Control settings: SUPPLY switch: STAND BY

1. Connect oscilloscope to 232C42 / 232L6.
2. Connect a shorting strap across capacitor 233 Cl .
3. Press TUNE button and adjust 232L6 for maximum deflection on oscilloscope.
4. Remove shorting strap.

### 7.11 /233\ Realignment of RF TRANSLATOR

Measuring equipment:

Oscilloscope having a sensitivity better than $50 \mathrm{mV} / \mathrm{div}$. Input impedance lomohms in parallel with 20 pF or less. Frequency range up to 50 MHz .

Standard Signal Generator covering the range $2-11 \mathrm{MHz}$ and having a frequency accuracy better than 10-3.

Receiver, heterodyne voltmeter or spectrum analyser covering the range $2-11 \mathrm{MHz}$.
7.11.1 Realignment of Level Setting potentiometers:

The Level Setting potentiometers control the RF drive voltage to the power output valves. Alignment procedure is given in section 7.8.3.
7.11.2 Realignment of 233 R4:

```
Control settings: SUPPLY switch: STANDBY
    MODE switch: A3H
    POWER switch: DUPLEX
    FREQUENCY NO: A radiotelephone frequency
    below 2800 kHz
```

1. Connect oscilloscope to output BNC socket 231SK7.
2. Adjust $233 R 4$ for minimum interference signal.
```
7.11.3 Realignment of 233R57:
    Control settings: SUPPLY switch: STAND BY
        MODE switch: A3H
        POWER switch: DUPLEX
        BAND switch: Not 2182 kHz
        FREQUENCY NO: A radiotelephone frequency in the
                            6 MHz band.
1. Connect oscilloscope to output BNC socket 231 SK7.
2. Adjust \(233 R 4\) tor minimum interference signa!.
7.11.4 Realignment of \(233 \mathrm{T1}:\)
Control settings: SUPPLY switch: STAND BY
1. Connect oscilloscope to 232C42 / 232L6.
2. Press TUNE button and adjust 233 T 1 tor minimum detlection on oscilloscope.
7.11.5 Realignment ot 4.3 MHz Low Pass Filter:
Contral settings: SUPPLY switch: STAND BY MODE switch: A3J
POWER switch: DUPLEX
BAND switch: Not 2182 kHz
FREQUENCY NO: Any radiotelephone frequency
1. Remove strap between terminals \(233-6\) and 233-7 and connect signal generator between same two terminals (233-7 is graund).
2. Connect receiver input via a coaxial cable to test point 233 TP 3 thraugh a 10 nF capacitor, and to ground at 233R16.
3. Set signal generator to \(4.6 \mathrm{MHz}, 100 \mathrm{mV}\).
4. Set receiver to same trequency and adjust tor a beat note.
5. Adjust 233L4 tor minimum signal (adjust receiver sensitivity to obtain a clear minimum).
6. Set signal generator and receiver to 4.7 MHz .
7. Adjust 233L5 tor minimum signal.
8. Set signal generator and receiver to 5.0 MHz.
9. Adjust 233L3 tor minimum signal.
10. Set signal generator and receiver to 6.14 MHz .
11. Adjust 233L6 tor minimum signal.
12. Set signal generator and receiver to 10.35 MHz .
13.Adjust 233L2 tor minimum signal.
14.Remove signal generator and receiver and strap terminals 233-6 and 233-7 together.
```

```
7.11.6 Realignment of 1.6-4 MHz filter:
    Control settings: SUPPLY switch: STAND BY
        MODE switch: A3J
        POWER switch: DUPLEX
        BAND switch: Not 2182 kHz
        FREQUENCY NO: A radiotelephone frequency between
                                1.6 and 4 MHz
1. Conneet signal generator to test point 233 TP 3 through a 10 nF capacitor and to ground at 233R16.
2. Connect receiver input to BNC socket 231SK7.
3. Set signal generator to \(4.43 \mathrm{MHz}, 100 \mathrm{mV}\).
4. Set receiver to same frequency and adjust for a beat note.
5. Adjust 233L22 for minimum signal (adjust receiver sensitivity to obtain a clear minimum).
6. Set signal generator and receiver to 4.83 MHz .
7. Adjust 233L24 for minimum signal.
8. Set signal generator and receiver to 7.13 MHz.
9. Adjust 233L20 for minimum signal.
\begin{tabular}{rl} 
7.11.7 & Realignment of Band Pass Filters \\
Control Settings: & SUPPLY switch: STAND BY \\
& \\
& MODE switch: \\
& POWER switch: DUPLEX \\
& BAND switch: \\
& Not 2182 kHz
\end{tabular}
1. Connect signal generator to test point 233 TP 3 through a 10 nF capacitor and to ground at 233R16.
2. Connect oscilloscope to BNC socket 231 SK7 through a coaxial cable terminated into 50 ohms at the oscilloscope.
3. Select by means of the keyboard a radiotelephone frequency in the band to be realigned.
4. Set signal generator to alignment frequency given in table below.
5. Connect a 1 kohm resistor between the alignment terminals across 233L32.
6. Adjust the input coil of the filter in question for maximum deflection on oscilloscope.
7. Move the 1 kohm resistor to the alignment terminals across 233R56.
8. Adjust the output coil of the filter for maximum deflection on the oscilloscope.
9. Remove 1 kohm resistor.
```

| Band | Alignment frequency <br> applied to 233TP3 | Input <br> Coil | Output <br> Coil |
| ---: | :---: | :---: | :---: |
| 4 MHz | 4140 kHz | 233 L 13 | 233 L 25 |
| 6 MHz | 2062 kHz | 233 L 14 | 233 L 26 |
| 8 MHz | 2715 kHz | 233 LI 5 | 233 L 27 |
| 12 MHz | 2690 kHz | 233 L 16 | 233 L 28 |
| 16 MHz | 2660 kHz | 233 L 17 | 233 L 29 |
| 22 MHz | 2560 kHz | 233 L 18 | 233 L 30 |
| 25 MHz | 2900 kHz | 233 L 19 | 233 L 31 |

7.12 /234\ Realignment of CONTROL CIRCUIT AND TONE GENERATORS

Measuring equipment:

Frequency Counter having an accuracy better than 10-4 and a sensitivity of at least 0.5 V .

Extender Board /259\.
7.12.1 Realignment of $234 \mathrm{T1}$ and 234 T 2 :

Control settings: SUPPLY switch: STAND BY MODE switch: TEST ALARM

1. Insert extender board.
2. Connect frequency counter between terminals 234-32c arid 234-4c (ground),
3. Connect the adjustment terminals marked 1 and 2 together.
4. Depress and release ALARM START pushbutton.
5. Adjust 234 T 1 until counter reads $2200 \mathrm{~Hz} \pm 1 \mathrm{~Hz}$.
6. Remove connection referred to in point 3 above.
7. Connect the terminals marked 2 and 3.
8. Depress and release ALARM START button.
9. Adjust 234 T 2 until counter reads $1300 \mathrm{~Hz} \pm 1 \mathrm{~Hz}$.
10. Remove connection referred to in point 7 above.
7.12.2 Realignment of 234T3:

Control settings: SUPPLY switch: STAND BY MODE switch: Al

1. Insert extender board,
2. Connect frequency counter between terminals 234-32e and 234-4e (ground),
3. Depress morse key or alternatively short circuit the terminals of the jack socket 230SK2.
4. Adjust 234 T 3 until counter reads $1500 \mathrm{~Hz} \pm 1 \mathrm{~Hz}$

## /235\ Realignment of HARMONIC GENERATOR

Measuring equipment:

DC Voltmeter having an internal resistance of 25 kohms or more.

Extender board /2591.

Control settings: SUPPLY switch: STAND BY
BAND switch: Not 2182 kHz

Board /235\ exists in two versions. The diagram no's. are 9922033 Version 2 and 9922033 Version 4. The printed circuit board of the latter version is labelled 4. Two different alignment procedures are applicable.

Version 2:

1. Insert extender board.
2. Connect voltmeter to test point 235 TP 1.
3. Connect pin no. 11 of $235 I C 7$ (74LS03) to ground.
4. Note voltmeter deflection and remove connection referred to in point 3.
5. Select a Frequency No. where no PROM's are installed on MEMORY board /238\.
6. Adjust 235 L2 for exactly the same deflection as in point 4 above.
7. Select by means of the Keyboard a frequency in the 25 MHz band and adjust 235L4 for the same deflection.
8. Select a frequency in the 22 MHz band and adjust 235 L 5 for the same deflection.
9. Select a frequency in the 16 MHz band and adjust 235 L 6 for the same deflection,
10. Select a frequency in the 12 MHz band and adjust 235 L 7 for the same deflection.
11. Select a frequency in the 8 MHz band and adjust 235 L 8 for the same deflection,
12. Select a frequency in the 6 MHz band and adjust 235L9 for the same deflection.

Version 4:

1. Insert extender board.
2. Connect voltmeter to test point 235 TP 4 .
3. Select a Frequency No. where no PROM's are installed on the MEMORY board $/ 238 \backslash$ and adjust $235 L 2 \mathrm{f}$ or 3.8 volts on the voltmeter.
4. Select by means of the Keyboard a frequency in the 25 MHz band and adjust 235L4 for 3.8 volts.
5. Select a frequency in the 22 MHz band and adjust 235 L 5 for 3.8 volts.
6. Select a frequency in the 16 MHz band and adjust 235 L 6 for 3.8 volts.
7. Select a frequency in the 12 MHz band and adjust 235 L 7 for 3.8 volts.
8. Select a frequency in the 8 MHz band and adjust 235 L 8 for 3.8 volts.
9. Select a frequency in the 6 MHz band and adjust 235 L 9 for 3.8 volts.

## /236\ Realignment of VCO Measuring equipment:

Frequency Counter having an accuracy better than 10-3 and a sensitivity of at least 0.5 V .

Extender Board /259\}
Control settings: SUPPLY switch: STAND BY
BAND switch: Not 2182 kHz

1. Insert extender board
2. Connect terminal 236-2a to terminal 236-4a (ground).
3. Connect frequency counter between terminal 236-30c and terminal 236$26 e$ (ground).
4. Select by means of the Keyboard a frequency in the range 1600 to 2599.9 kHz .
5. Adjust 236 T 1 until the counter reads 4.30 MHz .
6. Select a frequency in the range 2600 to 3599.9 kHz .
7. Adjust 236 L 2 until the counter reads 5.35 MHz .
8. Select a frequency in the 4 MHz band.
9. Adjust 236 L 3 until the counter reads 6.45 MHz .
10. Remove connection referred to under point 2 above.
7.15 /240\ Realignment of MASTER OSCILLATOR

Measuring equipment:

Oscilloscope having an input impedance of lOMohms in parallel with 20 PF or less.
Control settings: SUPPLY switch: STAND BY

1. Connect oscilloscope to terminals 240-2 and 240-3 (ground).
2. Adjust 240 T for maximum deflection on oscilloscope.

### 7.16 /251

Measuring equipment:

DC Voltmeter.

1. Disconnect HT-supply to Transmitter Power Amplifier by removing the plug to 250SK1.
2. Connect voltmeter to emitter of 251 TR3 or 251 TR4,
3. Set SUPPLY switch to TRANSMIT, select a Frequency No. and set BAND and MODE switches accordingly.
4. Key the transmitter and adjust 251 R 7 for 1.1 V on the voltmeter (if maximum voltage is below 1.1 V strap terminals 251-2 and 251-3 together etc.)

### 7.17 /254\ Realignment of VOLTAGE REGULATORS

Potentiometers 254 R 1 and 254 R 3 control the quiescent current of the power output valves. Alignment procedure is given in section 7.8.2.
001111 Transmit Alarm
$\begin{array}{lllll}0 & 1 & 0 & 0 & 0\end{array}$ A2H
$\begin{array}{lllll}0 & 1 & 0 & 0 & 1\end{array}$ A1

| 0 | 1 | 0 | 1 | 0 | $F 1$ |
| :--- | :--- | :--- | :--- | :--- | :--- |


| 0 | 1 | 0 | 1 | 1 |
| :--- | :--- | :--- | :--- | :--- | Test Alarm

$\begin{array}{lllll}0 & 1 & 1 & 0 & 0 \\ 0\end{array}$
$\begin{array}{lllll}0 & 1 & 1 & 0 & 1\end{array}$ A3J
$\begin{array}{lllll}0 & 1 & 1 & 1 & 0 \\ \text { A } 3 \mathrm{H}\end{array}$
$0 \begin{array}{lllll}0 & 1 & 1 & 1 & 1\end{array}$ Transmit Alarm

| 1 | 0 | 0 | 0 | 0 | $A 2 H$ |
| :--- | :--- | :--- | :--- | :--- | :--- |

$1 \begin{array}{lllll}1 & 0 & 0 & 0 & 1\end{array}$ A1

1 | 1 | 0 | 0 | 1 | 0 | $F 1$ |
| :--- | :--- | :--- | :--- | :--- | :--- |

1 | 1 | 0 | 0 | 1 | 1 |
| :--- | :--- | :--- | :--- | :--- | Test Alarm

$\begin{array}{lllll}1 & 0 & 1 & 0 & 0\end{array}$ A3A

| 1 | 0 | 1 | 0 | 1 | $A 3 J$ |
| :--- | :--- | :--- | :--- | :--- | :--- |


| 1 | 0 | 1 | 1 | 0 | $A 3 H$ |
| :--- | :--- | :--- | :--- | :--- | :--- |

$1 \begin{array}{lllll}1 & 0 & 1 & 1 & 1\end{array}$ Transmit Alarm
$\begin{array}{lllll}1 & 1 & 0 & 0 & 0 \\ \text { A2 } 2 H\end{array}$
$\begin{array}{lllll}1 & 1 & 0 & 0 & 1\end{array}$ A1

| 1 | 1 | 0 | 1 | 0 | $F 1$ |
| :--- | :--- | :--- | :--- | :--- | :--- |

1 | 1 | 1 | 0 | 1 | 1 | Test Alarm |
| :--- | :--- | :--- | :--- | :--- | :--- |

| 1 | 1 | 1 | 0 | 0 | $A 3 A$ |
| :--- | :--- | :--- | :--- | :--- | :--- |

$\begin{array}{lllll}1 & 1 & 1 & 0 & 1\end{array}$ A3J
$\begin{array}{lllll}1 & 1 & 1 & 1 & 0 \\ \text { A3H }\end{array}$

| 1 | 1 | 1 | 1 | 1 | Transmit Alarm |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Function table for $2341 C 5$
$32 x 8$ bit PROM of control circuit and tone generators

Function table for 235 IC4
$32 x 8$ bit PROM of Harmonic Generators

